

## Precision Low-voltage Amplifier; DC to 2 kHz

### Features

- Low Offset: 10  $\mu$ V Max
- Low Drift: 0.05  $\mu$ V/ $^{\circ}$ C Max
- Low Noise
  - 6 nV/ $\sqrt{\text{Hz}}$  @ 0.5 Hz
  - 0.1 to 10 Hz = 125 nVp-p
  - 1/f corner @ 0.08 Hz
- Open-loop Voltage Gain
  - 300 dB Typical
  - 200 dB Minimum
- Rail-to-rail Output Swing
- Slew Rate: 5 V/ $\mu$ s

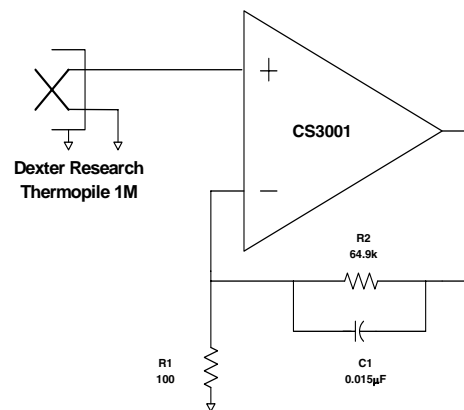
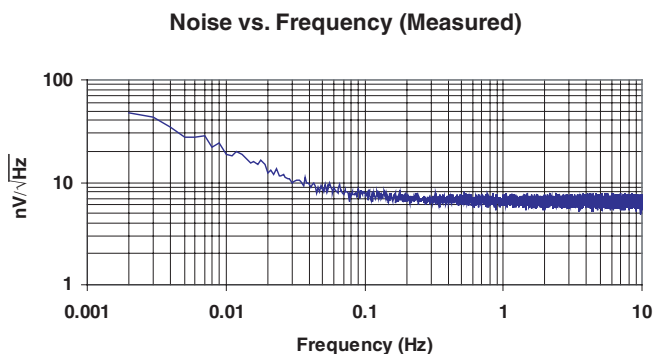
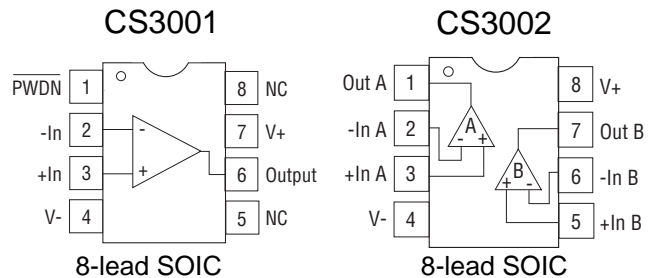
### Applications

- Thermocouple/Thermopile Amplifiers
- Load Cell and Bridge Transducer Amplifiers
- Precision Instrumentation
- Battery-powered Systems

### Description

The CS3001 single amplifier and the CS3002 dual amplifier are designed for precision amplification of low-level signals and are ideally suited to applications that require very high closed-loop gains. These amplifiers achieve excellent offset stability, super-high open-loop gain, and low noise over time and temperature. The devices also exhibit excellent CMRR and PSRR. The common mode input range includes the negative supply rail. The amplifiers operate with any total supply voltage from 2.7 V to 6.7 V ( $\pm 1.35$  V to  $\pm 3.35$  V).

### Pin Configurations



Thermopile Amplifier with a Gain of 650 V/V

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## Contacting Cirrus Logic Support

For all product questions and inquiries contact a Cirrus Logic Sales Representative.

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# 1. CHARACTERISTICS AND SPECIFICATIONS

## ELECTRICAL CHARACTERISTICS $V_+ = +5\text{ V}$ , $V_- = 0\text{ V}$ , $V_{CM} = 2.5\text{ V}$ (Note 1)

Parameter		CS3001/CS3002			Unit		
		Min	Typ	Max			
Input Offset Voltage	(Note 2)	•	-	-	±10	μV	
Average Input Offset Drift	(Note 2)	•	-	±0.01	±0.05	μV/°C	
Long Term Input Offset Voltage Stability			(Note 3)				
Input Bias Current	$T_A = 25^\circ\text{ C}$	•	-	±100	-	pA	
		•	-	-	±1000	pA	
Input Offset Current	$T_A = 25^\circ\text{ C}$	•	-	±200	-	pA	
		•	-	-	±2000	pA	
Input Noise Voltage Density	$R_S = 100\ \Omega$ , $f_0 = 1\ \text{Hz}$		-	6		nV/√Hz	
	$R_S = 100\ \Omega$ , $f_0 = 1\ \text{kHz}$		-	6		nV/√Hz	
Input Noise Voltage	0.1 to 10 Hz		-	125		nV <sub>p-p</sub>	
Input Noise Current Density	$f_0 = 1\ \text{Hz}$		-	100		fA/√Hz	
Input Noise Current	0.1 to 10 Hz		-	1.9		pA <sub>p-p</sub>	
Input Common Mode Voltage Range		•	-0.1	-	(V+)-1.25	V	
Common Mode Rejection Ratio (dc)	(Note 4)	•	115	120	-	dB	
Power Supply Rejection Ratio		•	120	136	-	dB	
Large Signal Voltage Gain	$R_L = 2\ \text{k}\Omega$ to $V_+/2$	(Note 5)	•	200	300	-	dB
Output Voltage Swing	$R_L = 2\ \text{k}\Omega$ to $V_+/2$	•	+4.7	-	-	V	
	$R_L = 100\ \text{k}\Omega$ to $V_+/2$			+4.99		V	
Slew Rate	$R_L = 2\ \text{k}\Omega$ , 100 pF			5	-	V/μs	
Overload Recovery Time			-	100	-	μs	
Supply Current	CS3001	•	-	2.1	2.8	mA	
	CS3002	•	-	3.6	4.8	mA	
	PWDN active (CS3001 Only)	(Note 6)	•		15	μA	
PWDN Threshold	(Note 6)	•	(V+) -1.0	-	-	V	
Start-up Time	(Note 7)	•	-	9	12	ms	

- Notes:
1. Symbol “•” denotes specification applies over -40 to +85 °C.
  2. This parameter is guaranteed by design and laboratory characterization. Thermocouple effects prohibit accurate measurement of these parameters in automatic test systems.
  3. 1000-hour life test data @ 125 °C indicates randomly distributed variation approximately equal to measurement repeatability of 1 μV.
  4. Measured within the specified common mode range limits.
  5. Guaranteed within the output limits of (V+ -0.3 V) to (V- +0.3 V). Tested with proprietary production test method.
  6. PWDN input has an internal pullup resistor to V+ of approximately 800 kΩ and is the major source of current consumption when PWDN is active low.
  7. The device has a controlled start-up behavior due to its complex open loop gain characteristics. Start-up time applies when supply voltage is applied or when PWDN is released.

## ABSOLUTE MAXIMUM RATINGS

Parameter	Min	Typ	Max	Unit
Supply Voltage [(V+) - (V-)]			6.8	V
Input Voltage	V- -0.3		V+ +0.3	V
Storage Temperature Range	-65		+150	°C

## 2. PERFORMANCE PLOTS

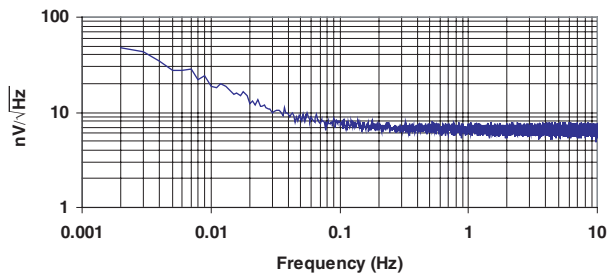


Figure 1. Noise vs Frequency (Measured)

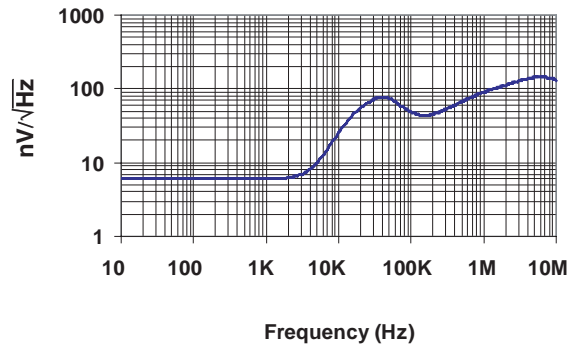


Figure 3. Noise vs Frequency

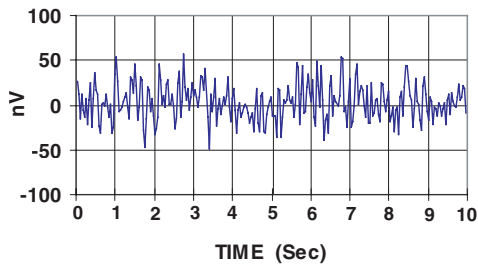


Figure 2. 0.01 Hz to 10 Hz Noise

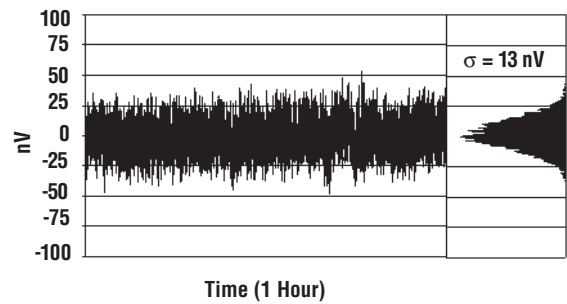
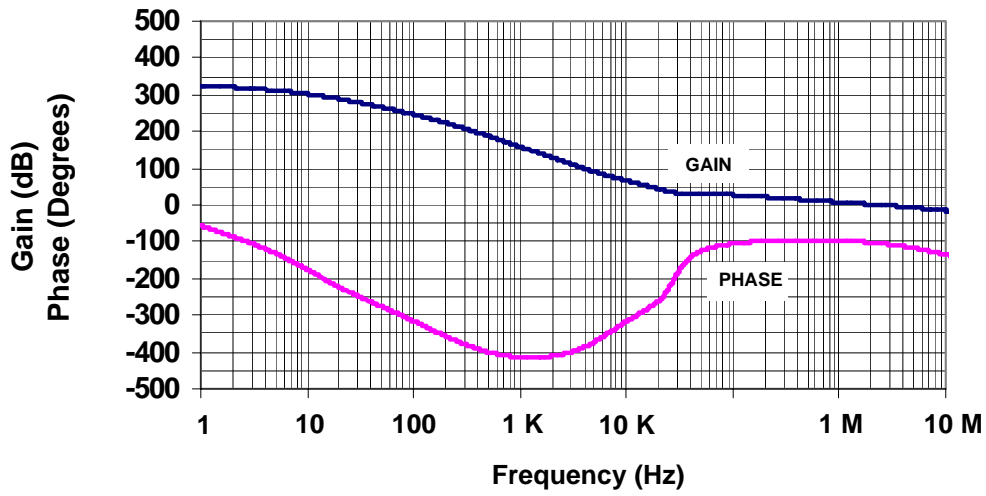
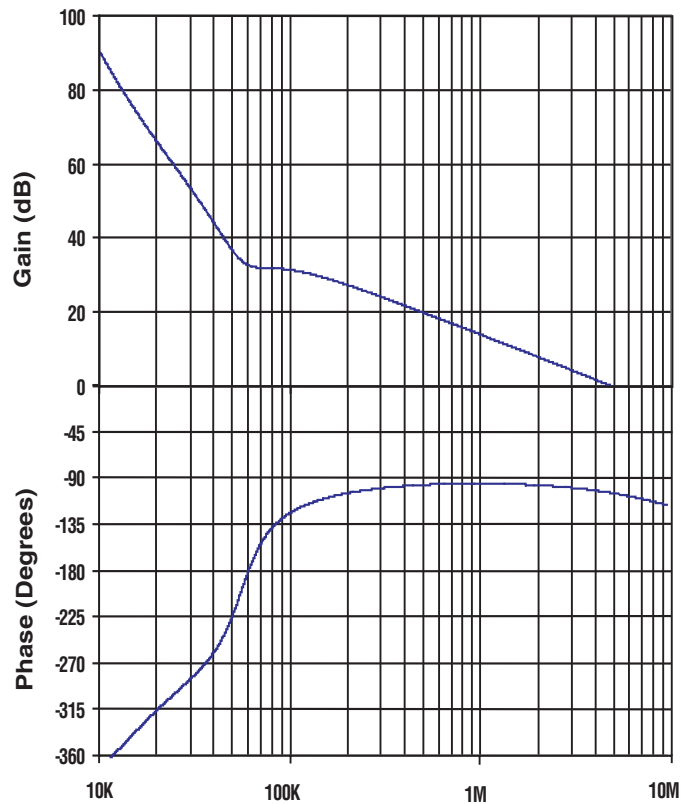
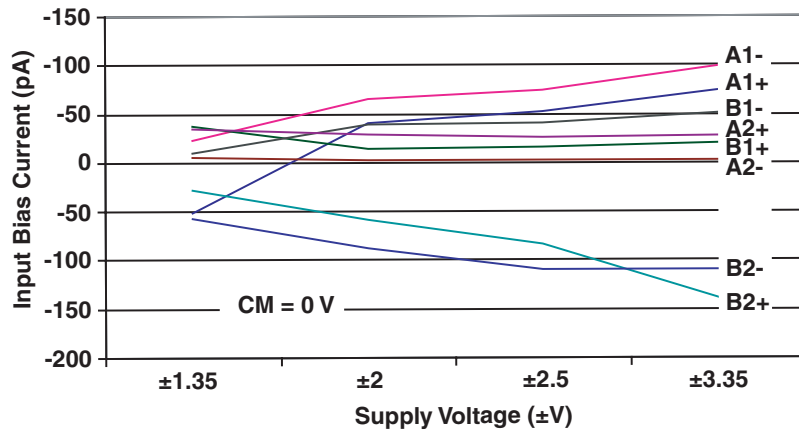
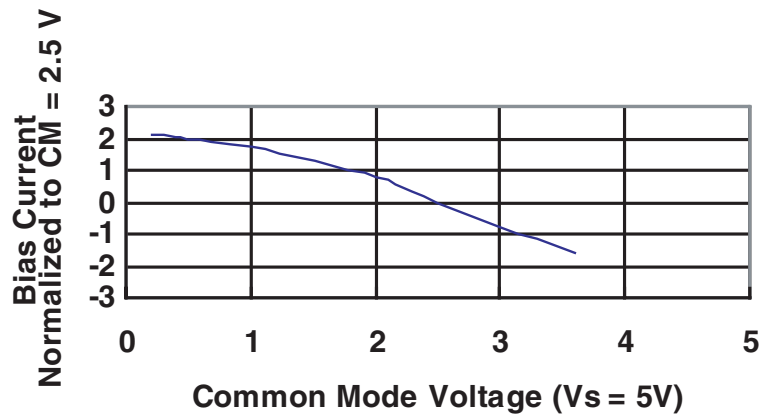


Figure 4. Offset Voltage Stability (DC to 3.2 Hz)

**Performance Plots (Cont.)**

**Figure 5. Open-loop Gain and Phase vs Frequency**

**Figure 6. Open-loop Gain and Phase vs Frequency (Expanded)**

**Performance Plots (Cont.)**

**Figure 7. Input Bias Current vs Supply Voltage (CS3002)**

**Figure 8. Input Bias Current vs Common Mode Voltage**

### 3. CS3001/CS3002 OVERVIEW

The CS3001/CS3002 amplifiers are designed for precision measurement of signals from DC to 2 kHz when operating from a supply voltage of +2.7 V to +6.7 V ( $\pm 1.35$  to  $\pm 3.35$  V). The amplifiers are designed with a patented architecture that utilizes multiple amplifier stages to yield very high open loop gain at frequencies of 10 kHz and below. The amplifiers yield low noise and low offset drift while consuming relatively low supply current. An increase in noise floor above 2 kHz is the result of intermediate stages of the amplifier being operated at very low currents. The amplifiers are intended for amplifying small signals with large gains in ap-

plications where the output of the amplifier can be band-limited to frequencies below 2 kHz.

#### 3.1 Open-loop Gain and Phase Response

Figure 9 illustrates the open loop gain and phase response of the CS3001/CS3002. The gain slope of the amplifier is about  $-100$  dB/decade between 500 Hz and 60 kHz and transitions to  $-20$  dB/decade between 60 kHz and its unity gain crossover frequency at about 4.8 MHz. Phase margin at unity gain is about 70 degrees; gain margin is about 20 dB.

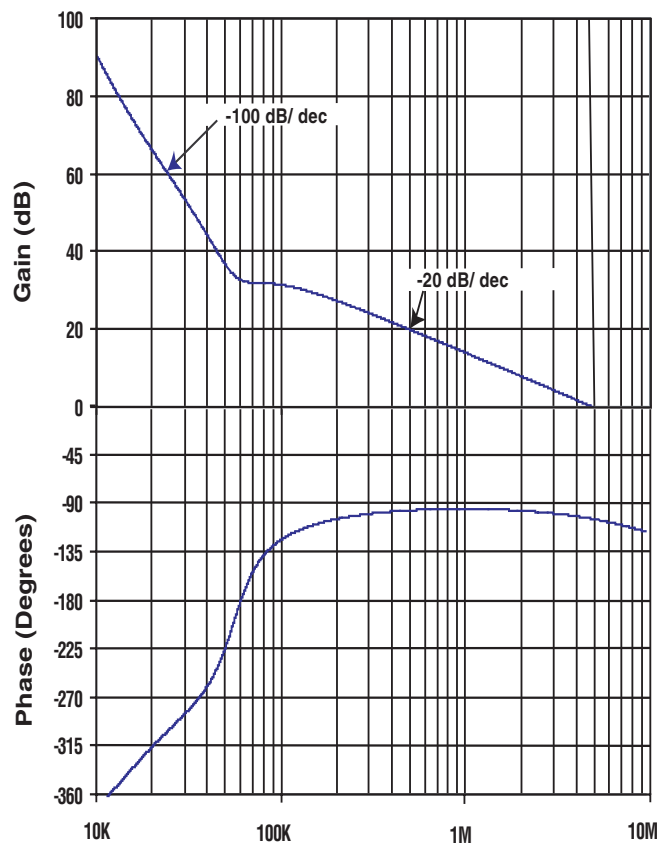


Figure 9. CS3001/CS3002 Open-loop Gain and Phase Response

### 3.2 Open-loop Gain and Stability Compensation

The CS3001 and CS3002 achieve ultra-high open loop gain. Figure 10 illustrates the amplifier in a non-inverting gain configuration. The open loop gain and phase plots indicate that the amplifier is stable for closed-loop gains less than 50 V/V. For a gain of 50, the phase margin is between 40° and 60° depending upon the loading conditions. As shown in Figure 11, on page 9, the operational amplifier has an input capacitance at the + and – signal inputs of typically 50 pF. This capacitance adds an addi-

tional pole in the loop gain transfer function at a frequency of  $f = 1/(2\pi R \cdot C_{in})$  where R is the parallel combination of R1 and R2 (R1 || R2). A higher value for R produces a pole at a lower frequency, thus reducing the phase margin. R1 is recommended to be less than or equal to 100 ohms, which results in a pole at 30 MHz or higher. If a higher value of R1 is desired, a compensation capacitor (C2) should be added in parallel with R2. C2 should be chosen such that  $R2 \cdot C2 \geq R1 \cdot C_{in}$ .

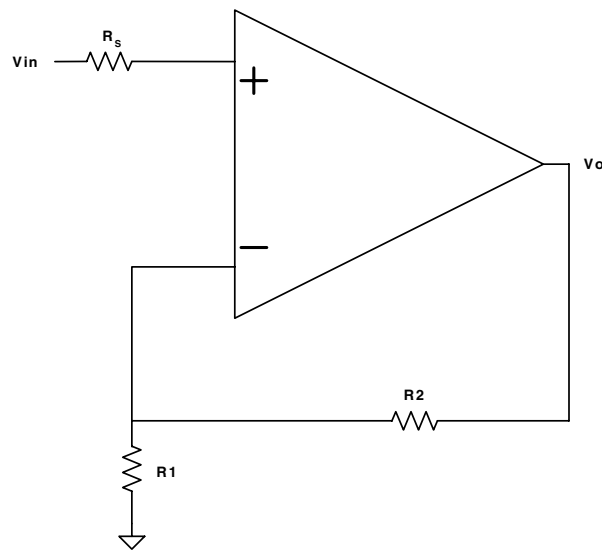
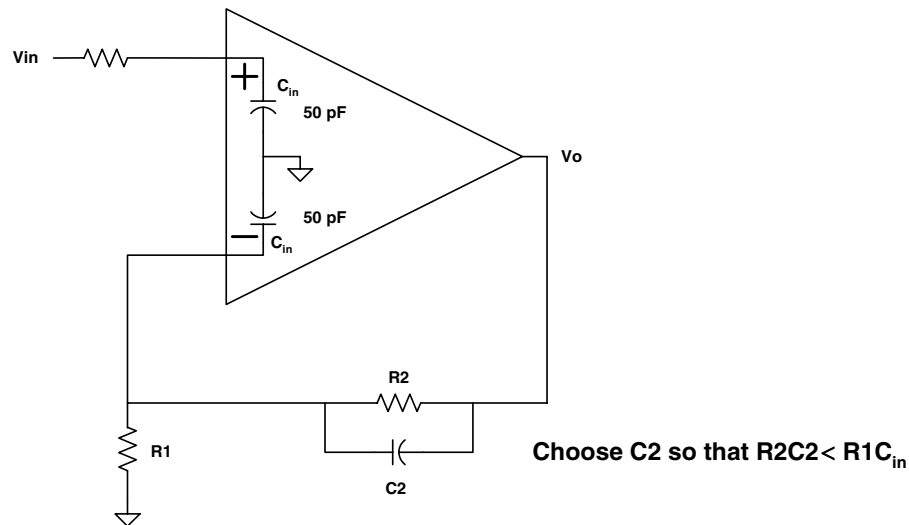


Figure 10. Non-inverting Gain Configuration





**Figure 11. Non-inverting Gain Configuration with Compensation**

The feedback capacitor C2 is required for closed-loop gains greater than 50 V/V. The capacitor introduces a pole and a zero in the loop gain transfer function,

$$T = \frac{-\left(1 + \frac{s}{z_1}\right)}{\left(1 + \frac{s}{p_1}\right)} A_{ol}$$

$$p_1 = \frac{1}{2\pi(R_1 \parallel R_2)C_2} \cong \frac{1}{2\pi(R_1 C_2)} \quad \text{for} \quad R_2 \gg R_1$$

$$z_1 = \frac{1}{2\pi(A \times R_1)C_2} \quad \text{where} \quad |A| = \frac{R_2}{R_1}$$

This indicates that the separation of the pole and the zero is governed by the closed loop gain. It is required that the zero falls on the steep slope ( $-100$  dB/decade) of the loop gain plot so that there is some gain higher than 0 dB (typically 20 dB) at the hand-over frequency (the frequency at which the slope changes from  $-100$  dB/decade to  $-20$  dB/decade).

The loop gain plot shown in Figure 12 illustrates the unity gain configuration, and indicates how this is modified when using the amplifier in a higher gain configuration with compensation. If it is configured for higher gain, for example, 60 dB, the x-axis will move up by 60 dB (line B). Capacitor C2 adds a zero and a pole. The modified plot indicates the effects of introducing the pole and zero due to capacitor C2. The pole can be located at any frequency higher than the hand-over frequency, the zero has to be at a frequency lower than the hand-over frequency so as to provide adequate gain mar-

gin. The separation between the pole and the zero is governed by the closed loop gain. The zero ( $z_1$ ) occurs at the intersection of the  $-100$  dB/decade and  $-80$  dB/decade slopes. The point X in the figure should be at closed loop gain plus 20 dB gain margin. The value for  $C_2 = 1/(2\pi R_1 p_1)$ . Using  $p_1 = 1$  MHz works very well and is independent of gain. As the closed loop gain is changed, the zero location is also modified if R1 remains fixed. Capacitor C2 can be increased in value to limit the amplifier's rising noise above 2 kHz.

### 3.3 Powerdown (PDWN)

The CS3001 single amplifier provides a power-down function on pin 1. If this pin is left open the amplifier will operate normally. If the powerdown is asserted low, the amplifier will go into a low power state. There is a pull-up resistor (approx-

imately 800 k ohm) inside the amplifier from pin 1 to the V+ supply. The current through this pull-up resistor is the main source of current drain in the powerdown state.

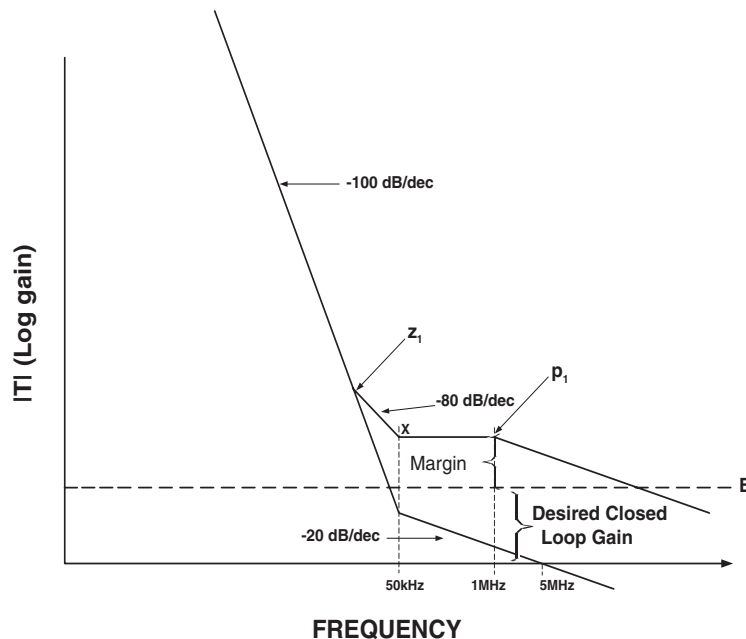


Figure 12. Loop Gain Plot: Unity Gain and with Pole-zero Compensation

### 3.4 Applications

The CS3001 and CS3002 amplifiers are optimum for applications that require high gain and low drift. Figure 13 illustrates a thermopile amplifier with a gain of 650 V/V. The thermopile outputs only a few millivolts when subjected to infrared radiation. The amplifier is compensated and bandlimited by C1 in combination with R2.

Figure 14, on page 11 illustrates a load cell bridge amplifier with a gain of 768 V/V. The load cell is

excited with +5 V and has a 1 mV/V sensitivity. Its full scale output signal is amplified to produce a fully differential  $\pm 3.8$  V into the CS5510/12 A/D converter. This circuit operates from +5 V.

A similar circuit operating from +3 V can be constructed using the CS5540/CS5541 A/D converters.

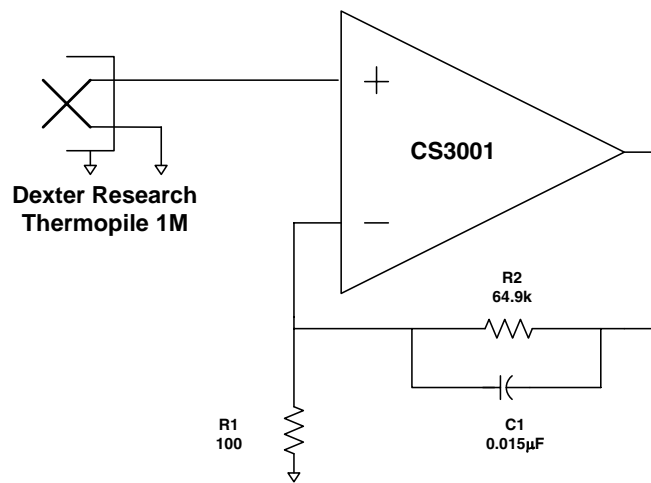
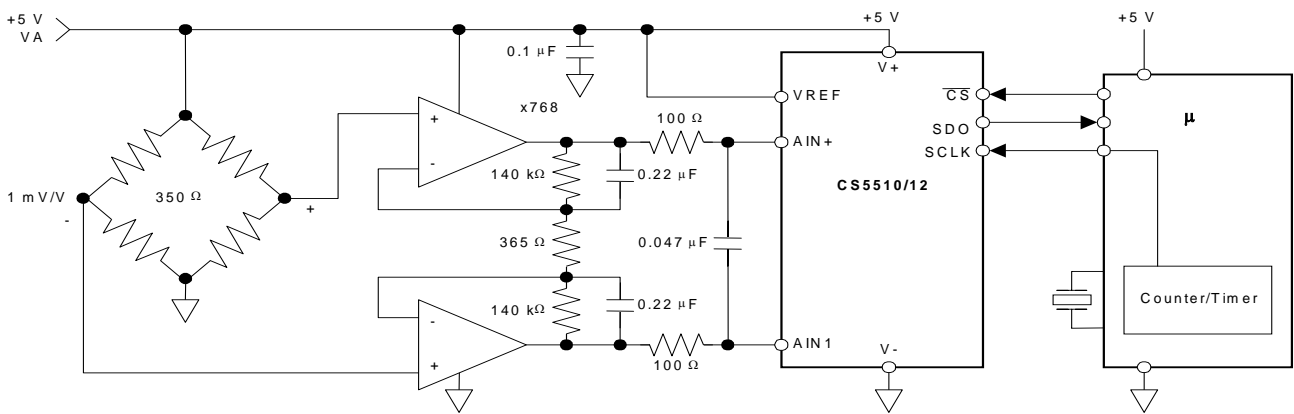


Figure 13. Thermopile Amplifier with a Gain of 650 V/V



SCLK = 10 kHz to 100 kHz  
(32.768 nominal)

Figure 14. Load Cell Bridge Amplifier and A/D Converter

#### 4. ORDERING INFORMATION

Model	Temperature	Package
CS3301-IS	-40 to +85 °C	8-pin SOIC
CS3301-ISZ (lead free)		
CS3302-IS		
CS3302-ISZ (lead free)		

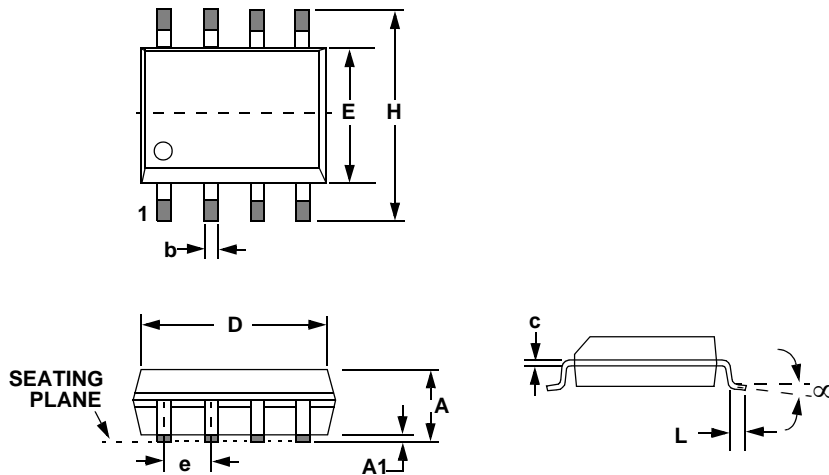
#### 5. ENVIRONMENTAL, MANUFACTURING, & HANDLING INFORMATION

Model Number	Peak Reflow Temp	MSL Rating*	Max Floor Life
CS3301-IS	240 °C	2	365 Days
CS3301-ISZ (lead free)	260 °C		
CS3302-IS	240 °C		
CS3302-ISZ (lead free)	260 °C		

\* MSL (Moisture Sensitivity Level) as specified by IPC/JEDEC J-STD-020.

#### 6. REVISION HISTORY

Revision	Date	Changes
F3	OCT 2004	Added lead-free device ordering information.
F4	August 2005	Added MSL specifications. Updated legal notice. Added leaded (Pb) devices.

**7. PACKAGE DRAWING**
**8L SOIC (150 MIL BODY) PACKAGE DRAWING**


DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.053	0.069	1.35	1.75
A1	0.004	0.010	0.10	0.25
B	0.013	0.020	0.33	0.51
C	0.007	0.010	0.19	0.25
D	0.189	0.197	4.80	5.00
E	0.150	0.157	3.80	4.00
e	0.040	0.060	1.02	1.52
H	0.228	0.244	5.80	6.20
L	0.016	0.050	0.40	1.27
∞	0°	8°	0°	8°

**JEDEC #: MS-012**

