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**Compact Video Driver Series for DSCs and Portable Devices** 

# Ultra-compact Waferlevel Chip Size Packeage Single Output Video Drivers



# BH76906GU, BH76909GU, BH76912GU, BH76916GU, BH76706GU

#### Description

Due to a built-in charge pump circuit, this video driver does not require the large capacity tantalum capacitor at the video output pin that is essential in conventional video drivers. Features such as a built-in LPF that has bands suited to mobile equipment, current consumption of 0  $\mu$ A at standby, and low voltage operation from as low as 2.5 V make it optimal for digital still cameras, mobile phones, and other equipment in which high density mounting is demanded.

#### Features

- 1) WLCSP ultra-compact package (1.6 mm x 1.6 mm x 0.75 mm)
- 2) Improved noise characteristics over BH768xxFVM series
- 3) Four video driver amplifier gains in lineup: 6 dB, 9 dB, 12 dB, 16.5 dB
- 4) Large output video driver of maximum output voltage 5.2 Vpp. Ample operation margin for supporting even low voltage operation
- 5) Output coupling capacitor not needed, contributing to compact design
- 6) Built-in standby function and circuit current of 0  $\mu$ A (typ) at standby
- 7) Clear image playback made possible by built-in 8<sup>th</sup>-order 4.5 MHz LPF
- 8) Due to use of bias input format, supports not only video signals but also chroma signals and RGB signals
- 9) Due to built-in output pin shunt switch, video output pin can be used as video input pin (BH76706GU)

### Application

Mobile phone, digital still camera, digital video camera, hand-held game, portable media player

	Product Name	Video Driver Amplifier Gain	Recommended Input Level	Video Output Pin Shunt Function
	BH76906GU	6dB	1Vpp	
	BH76909GU	9dB	0.7Vpp	
ſ	BH76912GU	12dB	0.5Vpp	_
	BH76916GU	16.5dB	0.3Vpp	-
	BH76706GU	6dB	1Vpp	0

●Lineup

• Absolute Maximum Ratings  $(T_a = 25 \degree C)$ 

Parameter	Symbol	Rating	Unit
Supply voltage	Vcc	3.55	V
Power dissipation	Pd	580	mW
Operating temperature range	Topr	-40~+85	°C
Storage temperature range	Tstg	-55~+125	°C

\* When mounted on a 50 mm×58 mm×1.6 mm glass epoxy board, reduce by 5.8mW/°C above Ta=+25°C.

# Operating Range

Parameter	Symbol	Min.	Тур.	Max.	Unit
Supply voltage	Vcc	2.5	3.0	3.45	V

## Electrical Characteristics

## [Unless otherwise specified, Ta = 25 °C, VCC = 3V]

Demonster	Typical Values			1.1	Magazinamant Canditiana			
Parameter	Symbol	76906	76909	76912	76916	76706	Unit	Measurement Conditions
Circuit current 1-1	CC1-1		15.0			mA	In active mode (No signal)	
Circuit current 1-2	I <sub>CC1-2</sub>		17.0			mA	In active mode (Outputting NTSC color bar signal)	
Circuit current 2	I <sub>CC2</sub>			0.0			μA	In standby mode
Circuit current 3	I <sub>CC3</sub>		-	_		100	μA	In input mode (Applying B3 = 1.5 V)
Standby switch input current High Level	I <sub>thH1</sub>		4	5			μA	Applying B3 = 3.0 V
Standby switch switching voltage High Level	$V_{\text{thH1}}$		1.2\	/ min		-	V	Active mode
Standby switch switching voltage Low Level	V <sub>thL1</sub>		0.45	Vmax			V	Standby mode
Standby switch outflow current High Level	I <sub>thH2</sub>					0	μA	Applying B3 = 3.0 V
Standby switch outflow current Middle Level	I <sub>thM2</sub>					8	μA	Applying B3 = 1.5 V
Standby switch outflow current Low Level	I <sub>thL2</sub>					23	μA	Applying B3 = 0 V
Mode switching voltage High Level	$V_{\text{thH2}}$	2		VCC -0.2 (MIN.)	V	Standby mode		
Mode switching voltage Middle Level	V <sub>thM2</sub>				VCC/2 (TYP.)	V	Input mode	
Mode switching voltage low Level	$V_{\text{thL2}}$					0.2 (MAX.)	V	Active mode
Voltage gain	Gv	6.0	9.0	12.0	16.5	6.0	dB	Vo=100kHz, 1.0Vpp
Maximum output level	Vomv			5.2			Vpp	f=10kHz,THD=1%
Frequency characteristic 1	G <sub>f1</sub>		-0.2		-0.2	dB	f=4.5MHz/100KHz	
Frequency characteristic 2	G <sub>f2</sub>		-1	.5		-1.4	dB	f=8.0MHz/100KHz
Frequency characteristic 3	G <sub>f3</sub>		-2	26		-28	dB	f=18MHz/100KHz
Frequency characteristic 4	G <sub>f4</sub>		_2	14		-48	dB	f=23.5MHz/100KHz
Differential gain	$D_G$			0.5			%	Vo=1.0Vp-p Inputting standard staircase Signal
Differential phase	D <sub>P</sub>			1.0			deg	Vo=1.0Vp-p Inputting standard staircase signal
Y signal to noise ratio	SNY	+74	+73	+70	+70	+74	dB	100 kHz~6MHz band Inputting 100% white video signal
C AM signal to noise ratio	SN <sub>CA</sub>	+77	+76	+75	+75	+77	dB	100~500 kHz band Inputting 100% chroma video signal
C PM signal to noise ratio	SN <sub>CP</sub>	+65			dB	100~500 kHz band Inputting 100% chroma video signal		
Current able to flow into output pin	lextin	30			mA	Applying 4.5 V to output pin through 150 Ω		
Output DC offset	Voff	±50max			mV	With no signal Voff = (Vout pin voltage) ÷ 2		
Input impedance	Rin			150			kΩ	Measure inflowing current when applying A3 = 1 V
Output pin shunt switch on resistance	Ron		-	_		3	Ω	



Fig. 1

※ A test circuit is a circuit for shipment inspection and differs from an application circuit example.

Block Diagram



(a) BH76906/09/12/16GU



(b) BH76706GU

Fig. 2

#### Operation Logic

BH769xxGU

STBY Pin Logic	Operating Mode	
Н	Active	
L	Standby	
OPEN	Standby	

#### BH76706GU

STBY Pin Logic	Operating Mode	SW1	SW2
Н	Standby	OFF	OFF
М	Input (Record)	ON	OFF
L	Active (Playback)	OFF	ON

%Use of the BH76706GU with the STBY pin OPEN is inappropriate

#### Pin Descriptions



Note 1) DC voltages in the figure are those when VCC = 3.0 V. Moreover, these values are reference values which are not guaranteed.

Note 2) Numeric values in the figure are settings which do not guarantee ratings.



Note 1) DC voltages in the figure are those when VCC = 3.0 V. Moreover, these values are reference values which are not guaranteed.

Note 2) Numeric values in the figure are settings which do not guarantee ratings.

#### Description of Operation

1) Principles of output coupling capacitorless video drivers



For an amplifier operated from a single power supply (single-supply), since the operating point has a potential of approximately 1/2 Vcc, a coupling capacitor is required for preventing direct current in the output. Moreover, since the load resistance is 150  $\Omega$  (75  $\Omega$  + 75  $\Omega$ ) for the video driver, the capacity of the coupling capacitor must be on the order of 1000  $\mu$ F if you take into account the low band passband. (Fig.3)

For an amplifier operated from dual power supplies (+ supply), since the operating point can be at GND level, a coupling capacitor for preventing output of direct current is not needed. Moreover, since a coupling capacitor is not needed, in principle, there is no lowering of the low band characteristic at the output stage. (Fig.4)

2) Occurrence of negative voltage due to charge pump circuit

A charge pump, as shown in Fig. 5, consists of a pair of switches (SW1, SW2) and a pair of capacitors (flying capacitor, anchor capacitor). Switching the pair of switches as shown in Fig. 5 causes a negative voltage to occur by shifting the charge in the flying capacitor to the anchor capacitor as in a bucket relay.

In this IC, by applying a voltage of +3 V, a negative voltage of approximately -2.8 V is obtained.



Fig.5 Principles of Charge Pump Circuit

#### 3) Configuration of BH769xxGU and BH76706GU

As shown in Fig. 6, a BH769xxGU or BH76706GU is a dual-supply amplifier and charge pump circuit integrated in one IC. Accordingly, while there is +3 V single-supply operation, since a dual-supply operation amplifier is used, an output coupling capacitor is not needed.



#### 4) Input pin format and sag characteristic

While a BH769xxGU or BH76706GU is a low voltage operation video driver, since it has a large dynamic range of approximately 5.2 Vpp, a resistance termination method that is compatible regardless of signal form (termination by 150 k $\Omega$ ) is used, and not a clamp method that is an input method exclusively for video signals.

Therefore, since a BH769xxGU or BH76706GU operates normally even if there is no synchronization signal in the input signal, it is compatible with not only normal video signals but also chroma signals and R.G.B. signals and has a wide application range.

Moreover, concerning sag (lowering of low band frequency) that occurs at the input pin and becomes a problem for the resistance termination method, since the input termination resistor is a high 150 k $\Omega$ , even if it is combined with a small capacity input capacitor, a sag characteristic that is not a problem in actual use is obtained.

In evaluating the sag characteristic, it is recommended that you use an H-bar signal in which sag readily stands out. (Fig. 8 to Fig. 10)



a) Video signal without sag (TG-7/1 output, H-bar)



Fig. 8

BH769xxGU or BH76706GU output (Input = 1.0 µF, TG-7/1 output, H-bar) b)







TG-7/1





Fig. 10

#### At playback (Active mode)



\* SW1 and SW2 are built-in BH76706GU only

#### Recording (Input mode) BH76706GU only



See page 3/16 for STBY pin logic in each mode

#### Fig.11

We are confident in recommending the above application circuit example, but we ask that you carefully check not just the static characteristics but also transient characteristics of this circuit before using it.

#### Caution on use

- Wiring from the decoupling capacitor C4 to the IC should be kept as short as possible. Moreover, this capacitor's capacitance value may have ripple effects on the IC, and may affect the S-N ratio for signals, so we recommend using as large a decoupling capacitor as possible. (Recommended C4: 3.3 μF, B characteristics, 6.3 V or higher maximum voltage) Make mount board patterns follow the layout example shown on page 10 as closely as possible.
- Capacitors to use In view of the temperature characteristics, etc., we recommend a ceramic capacitor with B characteristics.
- The NVCC (C1 pin) terminal generates a voltage that is used within the IC, so it should never be connected to a load unless absolutely necessary. Moreover, this capacitor (C2) has a large capacitance value but very little negative voltage ripple. (Recommended C2: 1.0 µF, B characteristic, 6.3 V or higher maximum voltage)

- Capacitors C1 and C4 should be placed as close as possible to the IC. If the wiring to the capacitor is too long, it can lead to intrusion of switching noise. (Recommended C1: 1.0 μF, B characteristics, 6.3 V or higher maximum voltage)
- 5. The HPF consists of input coupling capacitor C3 and 150 kΩ of internal input impedance. Be sure to check for video signal sag before determining the C3 value. The cut-off frequency fc can be calculated using the following formula. fc = 1/(2 π × C3 × 150kΩ) (Recommended C3: 1.0 μF, B characteristic, 6.3 V or higher maximum voltage)
- 6. The output resistor R2 should be placed close to the IC.
- 7. If the IC is mounted in the wrong direction, there is a risk of damage due to problems such as inverting VCC and GND. Be careful when mounting it.
- 8. A large current transition occurs in the power supply pin when the charge pump circuit is switched. If this affects other ICs (via the power supply line), insert a resistor (approximately 10  $\Omega$ ) in the VCC line to improve the power supply's ripple effects. Although inserting a 10  $\Omega$  resistor lowers the voltage by about 0.2 V, this IC has a wide margin for low-voltage operation, so dynamic range problems or other problems should not occur. (See Figures 12 to 14.)



Fig.12 Effects of Charge Pump Circuit Current Ripple on External Circuit





Waveform of current between power supply and capacitor (A) 10 mA/div

Waveform of current between capacitor and IC (B) 10 mA/div



Vcc

2) Decoupling capacitor + 10 Ω resistor







Vcc





Layer 1 wiring + Silkscreen legend

Layer 2 wiring

Solder pattern

Fig	•	1	5
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Parts List					
Symbol	Function	Recommended Value	Remarks		
C1	Flying capacitor	1μF	B characteristic recommended		
C2	Tank capacitor	1μF	B characteristic recommended		
C3	Input coupling capacitor	1μF	B characteristic recommended		
C4	Decoupling capacitor	3.3 µ F	B characteristic recommended		
R1	Input termination resistor	75Ω	Needed when connected to video signal measurement set		
R2	Output resistor	75Ω	—		
D2		75.0	Not needed when connected to TV or video signal		
КЭ	Output termination resistor	7502	measurement set		
	Input connector	BNC			
	Output connector	RCA (Pin jack)			

Reference Data



Fig. 16 Circuit Current vs Supply Voltage



Fig. 19 Standby Circuit Current vs Ambient Temperature



vs Supply Voltage





BH76906GU

VCC=3V



Fig. 23 VOUT Pin Output DC Offset vs Ambient Temperature





Ta=25℃

BH76906GU



Fig. 21 GND Mode Circuit Current vs Ambient Temperature



Fig. 24 Frequency Characteristic



Fig. 27 Voltage Gain vs Ambient Temperature



Fig. 28 Frequency Characteristic 1 vs Supply Voltage



Fig. 31 Frequency Characteristic 2 vs Ambient Temperature



POWER SUPPLY VOLTAGE [V]





Fig. 37 Max. Output Level vs Ambient Temperature



Fig. 29 Frequency Characteristic 1 vs Ambient Temperature



POWER SUPPLY VOLTAGE [V]

Fig.32 Frequency Characteristic 3 vs Supply Voltage



TEMPERATURE [℃]

Fig. 35 Frequency Characteristic 4 vs Ambient Temperature





Fig. 30 Frequency Characteristic 2 vs Supply Voltage



TEMPERATURE [°C] Fig.33 Frequency Characteristic 3 vs Ambient Temperature



POWER SUPPLY VOLTAGE [V]

Fig. 36 Max. Output Level vs Supply Voltage

![](_page_12_Figure_0.jpeg)

![](_page_12_Figure_1.jpeg)

![](_page_12_Figure_2.jpeg)

![](_page_13_Figure_0.jpeg)

Fig. 47 Y S/N vs Supply Voltage

![](_page_13_Figure_2.jpeg)

Fig. 49 C AM S/N vs Supply Voltage

![](_page_13_Figure_4.jpeg)

Fig. 51 C PM S/N vs Supply Voltage

![](_page_13_Figure_6.jpeg)

Fig. 53 Input Impedance vs Supply Voltage

![](_page_13_Figure_8.jpeg)

TEMPERATURE [°C] Fig.48 Y S/N vs Ambient Temperature

![](_page_13_Figure_10.jpeg)

Fig. 50 C AM S/N vs Ambient Temperature

![](_page_13_Figure_12.jpeg)

Fig. 52 C PM S/N vs Ambient Temperature

![](_page_13_Figure_14.jpeg)

![](_page_13_Figure_15.jpeg)

![](_page_14_Figure_0.jpeg)

Fig. 55 Control Pin Characteristic

![](_page_14_Figure_2.jpeg)

Fig. 57 Output Pin Shunt Switch On Resistance vs Supply Voltage

![](_page_14_Figure_4.jpeg)

Fig. 56 Control Pin Characteristic

![](_page_14_Figure_6.jpeg)

Fig. 58 Output Pin Shunt Switch On Resistance vs Ambient Temperature

#### Performing separate electrostatic damage countermeasures

When adding an externally attached electrostatic countermeasure element to the output pin, connect a varistor in the position shown in Fig. 59 (if connected directly to the output pin, the IC could oscillate depending on the capacity of the varistor). For this IC, since the output waveform is GND-referenced and swings positive and negative, a normal Zener diode cannot be used.

![](_page_14_Figure_10.jpeg)

Fig.59 Using Externally Attached Varistor

![](_page_15_Figure_1.jpeg)

Tape and Reel information

2

![](_page_15_Figure_3.jpeg)

#### VCSP85H1

![](_page_15_Figure_5.jpeg)

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![](_page_15_Picture_18.jpeg)

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Appendix1-Rev2.0

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