- **33-mΩ (5-V Input) High-Side MOSFET Switch**
- **Short-Circuit and Thermal Protection**
- **Overcurrent Logic Output**
- **Operating Range . . . 2.7 V to 5.5 V**
- **Logic-Level Enable Input**
- **Typical Rise Time . . . 6.1 ms**
- **Undervoltage Lockout**
- **Maximum Standby Supply Current . . . 10 μA**
- **No Drain-Source Back-Gate Diode**
- **Available in 8-pin SOIC and PDIP Packages**
- **Ambient Temperature Range, –40°C to 85°C**
- **2-kV Human-Body-Model, 200-V Machine-Model ESD Protection**
- **UL Listed – File No. E169910**

D OR P PACKAGE
(TOP VIEW)

GND	1	8	OUT
IN	2	7	OUT
IN	3	6	OUT
$\overline{EN}$	4	5	$\overline{OC}$
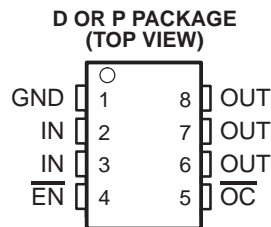
description

The TPS202x family of power distribution switches is intended for applications where heavy capacitive loads and short circuits are likely to be encountered. These devices are 50-mΩ N-channel MOSFET high-side power switches. The switch is controlled by a logic enable compatible with 5-V logic and 3-V logic. Gate drive is provided by an internal charge pump designed to control the power-switch rise times and fall times to minimize current surges during switching. The charge pump requires no external components and allows operation from supplies as low as 2.7 V.

When the output load exceeds the current-limit threshold or a short is present, the TPS202x limits the output current to a safe level by switching into a constant-current mode, pulling the overcurrent ($\overline{OC}$) logic output low. When continuous heavy overloads and short circuits increase the power dissipation in the switch, causing the junction temperature to rise, a thermal protection circuit shuts off the switch to prevent damage. Recovery from a thermal shutdown is automatic once the device has cooled sufficiently. Internal circuitry ensures the switch remains off until valid input voltage is present.

The TPS202x devices differ only in short-circuit current threshold. The TPS2020 limits at 0.3-A load, the TPS2021 at 0.9-A load, the TPS2022 at 1.5-A load, the TPS2023 at 2.2-A load, and the TPS2024 at 3-A load (see Available Options). The TPS202x is available in an 8-pin small-outline integrated-circuit (SOIC) package and in an 8-pin dual-in-line (DIP) package and operates over a junction temperature range of –40°C to 125°C.

GENERAL SWITCH CATALOG									
33 mΩ, single	TPS201xA	0.2 A – 2 A	**80 mΩ, dual**	TPS2042	500 mA	**80 mΩ, triple**		**80 mΩ, quad**	
	TPS202x	**0.2 A – 2 A**		TPS2052	500 mA				
	TPS203x	0.2 A – 2 A		TPS2046	250 mA				
				TPS2056	250 mA				
80 mΩ, single	TPS2014	600 mA	**260 mΩ**	TPS2100/1		TPS2043	500 mA	TPS2044	500 mA
	TPS2015	1 A	IN1	IN1	500 mA	TPS2053	500 mA	TPS2054	500 mA
	TPS2041	500 mA	IN2	IN2	10 mA	TPS2047	250 mA	TPS2048	250 mA
	TPS2051	500 mA		TPS2102/3/4/5		TPS2057	250 mA	TPS2058	250 mA
	TPS2045	250 mA	**1.3 Ω**	IN1	500 mA				
	TPS2055	250 mA		IN2	100 mA				

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

TEXAS INSTRUMENTS

TPS2020, TPS2021, TPS2022, TPS2023, TPS2024
POWER-DISTRIBUTION SWITCHES

AVAILABLE OPTIONS

T_A	ENABLE	RECOMMENDED MAXIMUM CONTINUOUS LOAD CURRENT (A)	TYPICAL SHORT-CIRCUIT CURRENT LIMIT AT 25°C (A)	PACKAGED DEVICES	
				SMALL OUTLINE (D)[†]	PLASTIC DIP (P)
−40°C to 85°C	Active low	0.2	0.3	TPS2020D	TPS2020P
		0.6	0.9	TPS2021D	TPS2021P
		1	1.5	TPS2022D	TPS2022P
		1.5	2.2	TPS2023D	TPS2023P
		2	3	TPS2024D	TPS2024P

[†] The D package is available taped and reeled. Add an R suffix to device type (e.g., TPS2020DR)

TPS2020 functional block diagram

†Current Sense
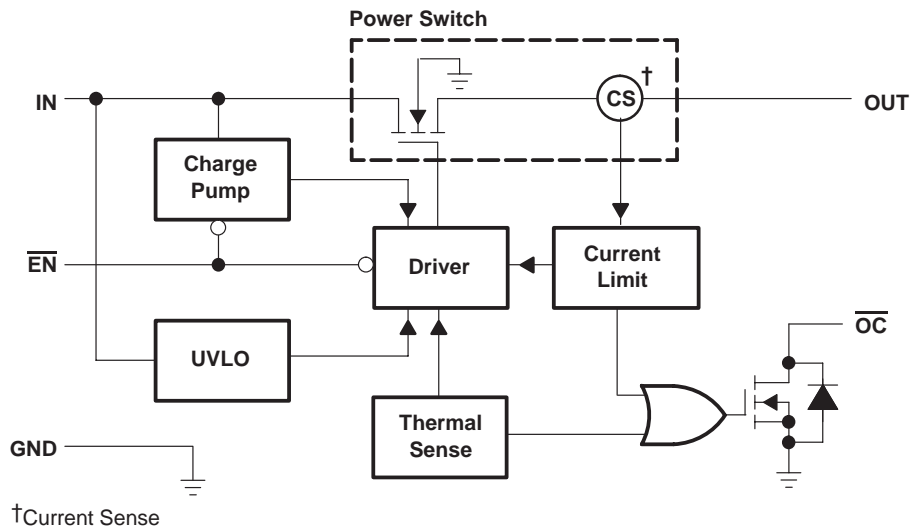
Terminal Functions

TERMINAL		I/O	DESCRIPTION
NAME	NO. D OR P		
$\overline{EN}$	4	I	Enable input. Logic low turns on power switch.
GND	1	I	Ground
IN	2, 3	I	Input voltage
$\overline{OC}$	5	O	Overcurrent. Logic output active low
OUT	6, 7, 8	O	Power-switch output

detailed description

power switch

The power switch is an N-channel MOSFET with a maximum on-state resistance of 50 mΩ ($V_{I(IN)}$ = 5 V). Configured as a high-side switch, the power switch prevents current flow from OUT to IN and IN to OUT when disabled.

charge pump

An internal charge pump supplies power to the driver circuit and provides the necessary voltage to pull the gate of the MOSFET above the source. The charge pump operates from input voltages as low as 2.7 V and requires very little supply current.

driver

The driver controls the gate voltage of the power switch. To limit large current surges and reduce the associated electromagnetic interference (EMI) produced, the driver incorporates circuitry that controls the rise times and fall times of the output voltage. The rise and fall times are typically in the 2-ms to 9-ms range.

enable ($\overline{\text{EN}}$)

The logic enable disables the power switch, the bias for the charge pump, driver, and other circuitry to reduce the supply current to less than 10 µA when a logic high is present on $\overline{\text{EN}}$. A logic zero input on $\overline{\text{EN}}$ restores bias to the drive and control circuits and turns the power on. The enable input is compatible with both TTL and CMOS logic levels.

overcurrent ($\overline{\text{OC}}$)

The $\overline{\text{OC}}$ open drain output is asserted (active low) when an overcurrent or overtemperature condition is encountered. The output will remain asserted until the overcurrent or overtemperature condition is removed.

current sense

A sense FET monitors the current supplied to the load. The sense FET measures current more efficiently than conventional resistance methods. When an overload or short circuit is encountered, the current-sense circuitry sends a control signal to the driver. The driver, in turn, reduces the gate voltage and drives the power FET into its saturation region, which switches the output into a constant current mode and holds the current constant while varying the voltage on the load.

thermal sense

An internal thermal-sense circuit shuts off the power switch when the junction temperature rises to approximately 140°C. Hysteresis is built into the thermal sense circuit. After the device has cooled approximately 20°C, the switch turns back on. The switch continues to cycle off and on until the fault is removed.

undervoltage lockout

A voltage sense circuit monitors the input voltage. When the input voltage is below approximately 2 V, a control signal turns off the power switch.

absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Input voltage range, $V_{I(IN)}$ (see Note 1) . . . . −0.3 V to 6 V
Output voltage range, $V_{O(OUT)}$ (see Note 1) . . . . −0.3 V to $V_{I(IN)}$ + 0.3 V
Input voltage range, $V_{I(\overline{EN})}$ . . . . −0.3 V to 6 V
Continuous output current, $I_{O(OUT)}$ . . . . internally limited
Continuous total power dissipation . . . . See Dissipation Rating Table
Operating virtual junction temperature range, T_J . . . . −40°C to 125°C
Storage temperature range, T_{stg} . . . . −65°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds . . . . 260°C
Electrostatic discharge (ESD) protection: Human body model . . . . 2 kV
                                            Machine model 200V
                                            Charged device model (CDM) . . . . 750 V

† Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and
   functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not
   implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
NOTE 1:   All voltages are with respect to GND.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25°C$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25°C$	$T_A = 70°C$ POWER RATING	$T_A = 85°C$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW
P	1175 mW	9.4 mW/°C	752 mW	611 mW

recommended operating conditions

		MIN	MAX	UNIT
Input voltage	$V_{I(IN)}$	2.7	5.5	V
	$V_{I(\overline{EN})}$	0	5.5	V
Continuous output current, I_O	TPS2020	0	0.2	A
	TPS2021	0	0.6	
	TPS2022	0	1	
	TPS2023	0	1.5	
	TPS2024	0	2	
Operating virtual junction temperature, T_J		−40	125	°C

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electrical characteristics over recommended operating junction temperature range, $V_{I(IN)}$= 5.5 V, I_O = rated current, $\overline{EN}$ = 0 V (unless otherwise noted)

power switch

PARAMETER		TEST CONDITIONS†			MIN	TYP	MAX	UNIT
$r_{DS(on)}$	Static drain-source on-state resistance	$V_{I(IN)}$ = 5 V,	T_J = 25°C,	I_O = 1.8 A		33	36	mΩ
		$V_{I(IN)}$ = 5 V,	T_J = 85°C,	I_O = 1.8 A		38	46	
		$V_{I(IN)}$ = 5 V,	T_J = 125°C,	I_O = 1.8 A		44	50	
		$V_{I(IN)}$ = 3.3 V,	T_J = 25°C,	I_O = 1.8 A		37	41	
		$V_{I(IN)}$ = 3.3 V,	T_J = 85°C,	I_O = 1.8 A		43	52	
		$V_{I(IN)}$ = 3.3 V,	T_J = 125°C,	I_O = 1.8 A		51	61	
		$V_{I(IN)}$ = 5 V,	T_J = 25°C,	I_O = 0.18 A		30	34	
		$V_{I(IN)}$ = 5 V,	T_J = 85°C,	I_O = 0.18 A		35	41	
		$V_{I(IN)}$ = 5 V,	T_J = 125°C,	I_O = 0.18 A		39	47	
		$V_{I(IN)}$ = 3.3 V,	T_J = 25°C,	I_O = 0.18 A		33	37	
		$V_{I(IN)}$ = 3.3 V,	T_J = 85°C,	I_O = 0.18 A		39	46	
		$V_{I(IN)}$ = 3.3 V,	T_J = 125°C,	I_O = 0.18 A		44	56	
t_r	Rise time, output	$V_{I(IN)}$ = 5.5 V, C_L = 1 µF,	T_J = 25°C, R_L = 10 Ω			6.1		ms
		$V_{I(IN)}$ = 2.7 V, C_L = 1 µF,	T_J = 25°C, R_L = 10 Ω			8.6		
t_f	Fall time, output	$V_{I(IN)}$ = 5.5 V, C_L = 1 µF,	T_J = 25°C, R_L = 10 Ω			3.4		ms
		$V_{I(IN)}$ = 2.7 V, C_L = 1 µF,	T_J = 25°C, R_L = 10 Ω			3		

† Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

enable input ($\overline{EN}$)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{IH}	High-level input voltage	2.7 V ≤ $V_{I(IN)}$ ≤ 5.5 V	2			V
V_{IL}	Low-level input voltage	4.5 V ≤ $V_{I(IN)}$ ≤ 5.5 V			0.8	V
		2.7 V ≤ $V_{I(IN)}$ ≤ 4.5 V			0.5	
I_I	Input current	$\overline{EN}$= 0 V or $\overline{EN}$ = $V_{I(IN)}$	−0.5		0.5	µA
t_{on}	Turnon time	C_L = 100 µF, R_L = 10 Ω			20	ms
t_{off}	Turnoff time	C_L = 100 µF, R_L = 10 Ω			40	

current limit

PARAMETER		TEST CONDITIONS†		MIN	TYP	MAX	UNIT
I_{OS}	Short-circuit output current	T_J = 25°C, V_I = 5.5 V, OUT connected to GND, Device enable into short circuit	TPS2020	0.22	0.3	0.4	A
			TPS2021	0.66	0.9	1.1	
			TPS2022	1.1	1.5	1.8	
			TPS2023	1.65	2.2	2.7	
			TPS2024	2.2	3	3.8	

† Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

TEXAS INSTRUMENTS

electrical characteristics over recommended operating junction temperature range, $V_{I(IN)}$= 5.5 V, I_O = rated current, $\overline{EN}$ = 0 V (unless otherwise noted) (continued)

supply current

PARAMETER	TEST CONDITIONS			MIN	TYP	MAX	UNIT
Supply current, low-level output	No Load on OUT	$\overline{EN}$ = $V_{I(IN)}$	T_J = 25°C		0.3	1	µA
			$-40°C \le T_J \le 125°C$			10	
Supply current, high-level output	No Load on OUT	$\overline{EN}$ = 0 V	T_J = 25°C		58	75	µA
			$-40°C \le T_J \le 125°C$		75	100	
Leakage current	OUT connected to ground	$\overline{EN}$ = $V_{I(IN)}$	$-40°C \le T_J \le 125°C$		10		µA

undervoltage lockout

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Low-level input voltage		2		2.5	V
Hysteresis	T_J = 25°C		100		mV

overcurrent ($\overline{OC}$)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output low voltage	I_O = 10 mA, $V_{OL(\overline{OC})}$			0.4	V
Off-state current[†]	V_O = 5 V, V_O = 3.3 V			1	µA

[†] Specified by design, not production tested.

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PARAMETER MEASUREMENT INFORMATION
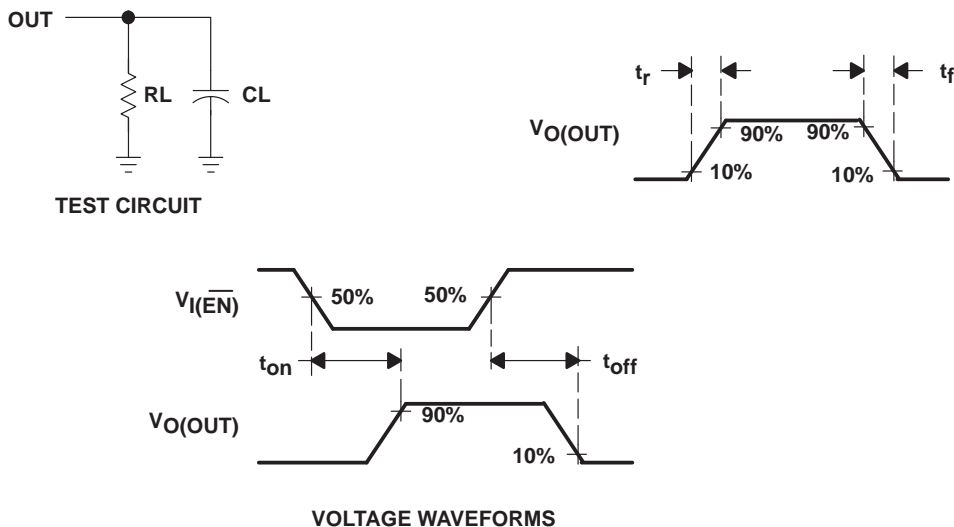
Figure 1. Test Circuit and Voltage Waveforms

Table of Timing Diagrams

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Turnon Delay and Rise TIme	2
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Turnoff Delay and Rise TIme with 1-μF Load	5
Device Enabled into Short	6
TPS2020, TPS2021, TPS2022, TPS2023, and TPS2024, Ramped Load on Enabled Device	7, 8, 9, 10, 11
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0.9-Ω Load Connected to an Enabled TPS2024 Device	21
0.5-Ω Load Connected to an Enabled TPS2024 Device	22

PARAMETER MEASUREMENT INFORMATION
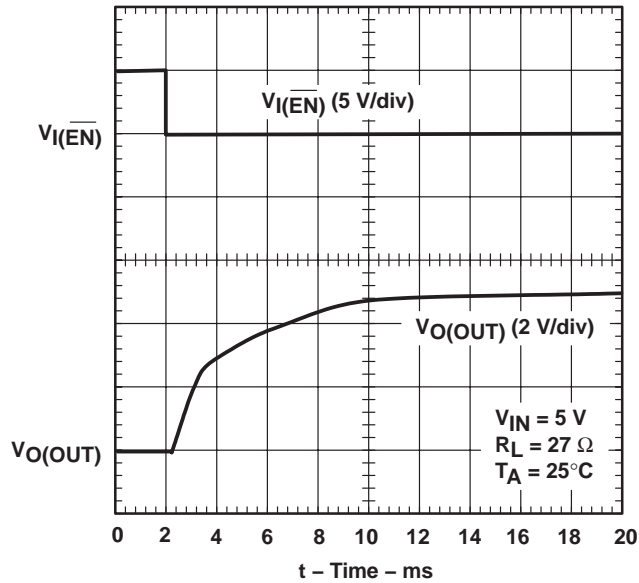
Figure 2. Turnon Delay and Rise Time
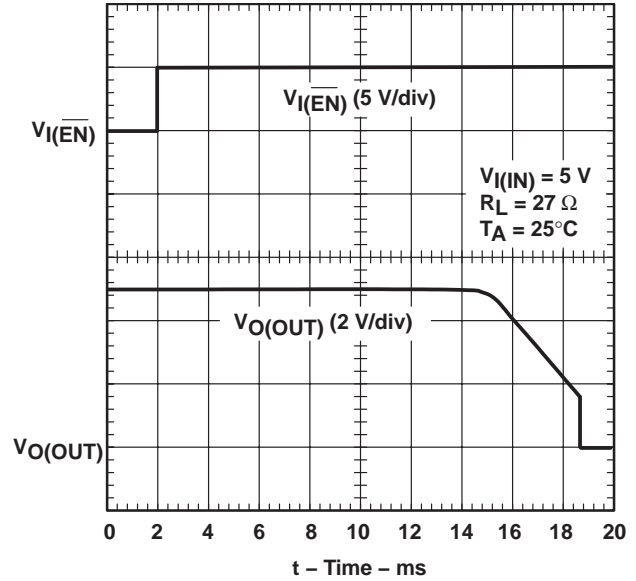
Figure 3. Turnoff Delay and Fall Time
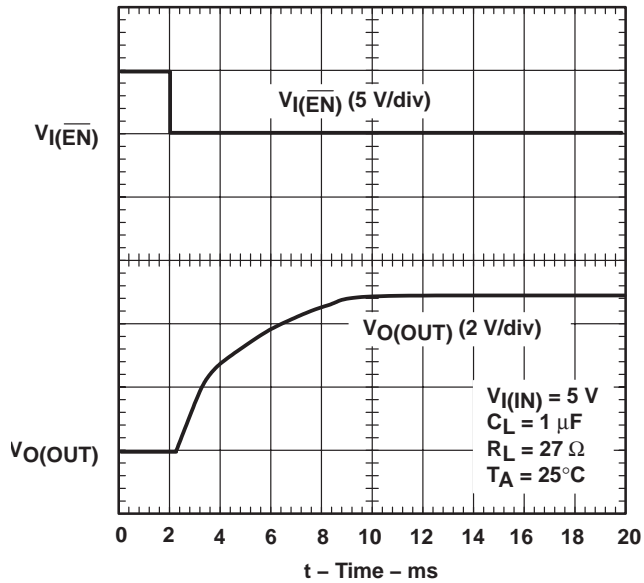
Figure 4. Turnon Delay and Rise Time
With 1-μF Load
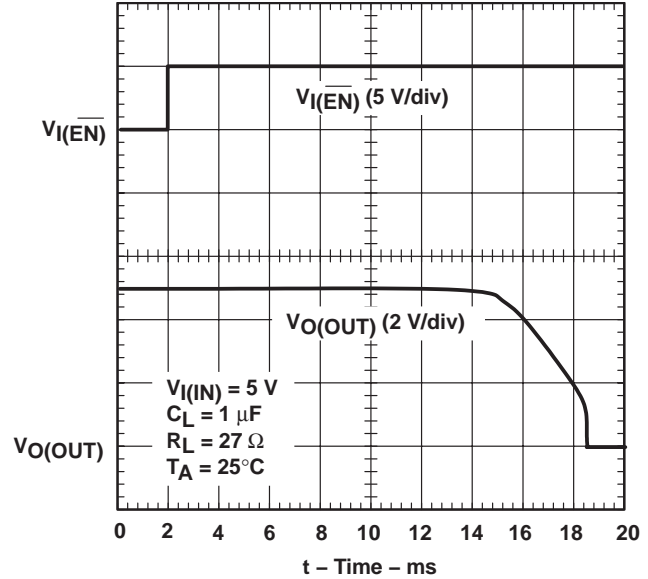
Figure 5. Turnoff Delay and Fall Time
With 1-μF Load

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PARAMETER MEASUREMENT INFORMATION
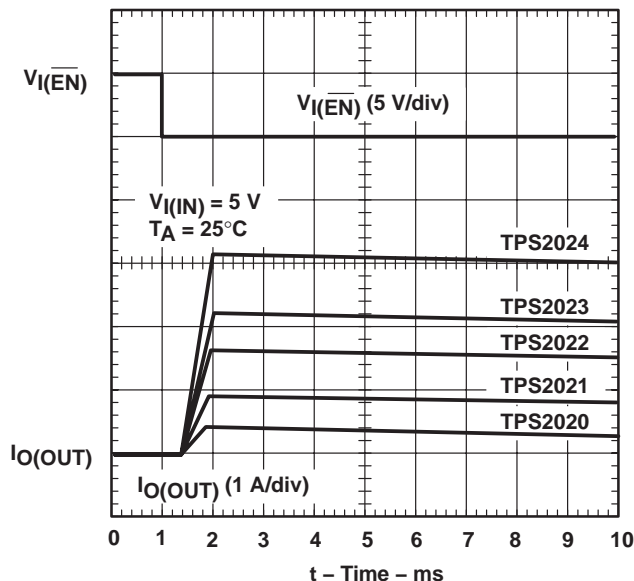
Figure 6. Device Enabled Into Short
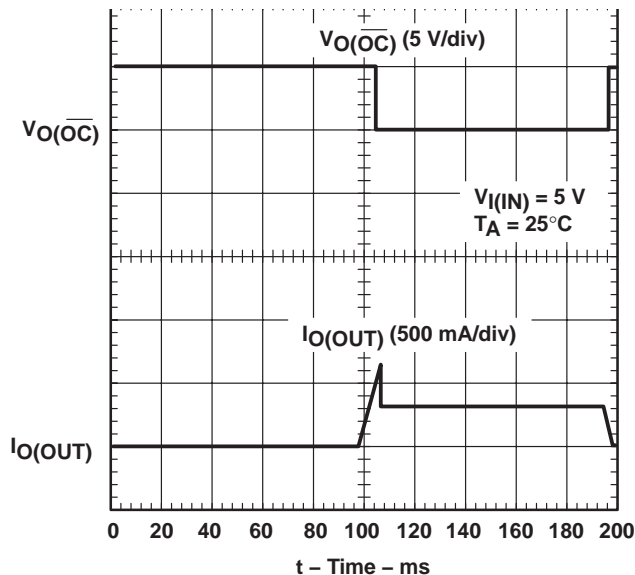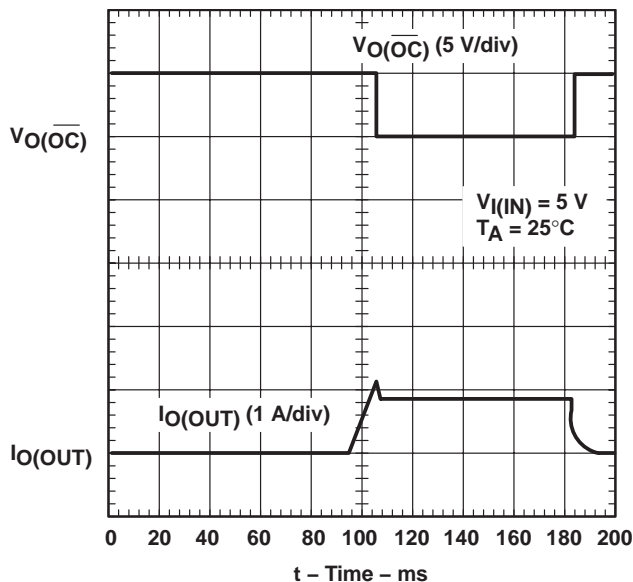

Figure 7. TPS2020, Ramped Load on Enabled Device
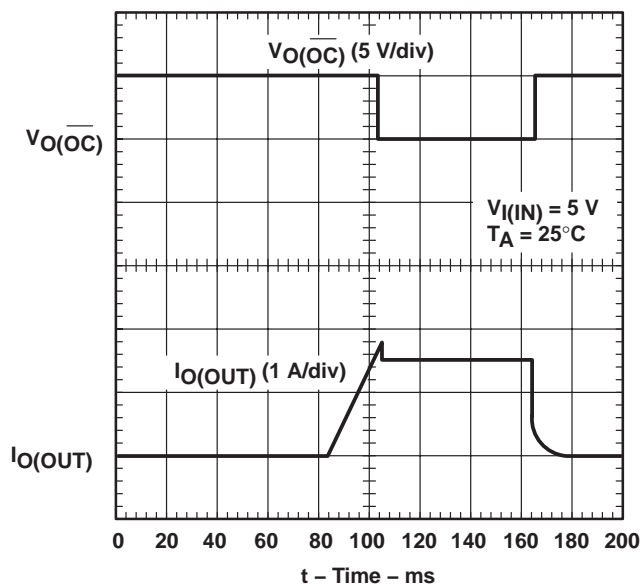

Figure 8. TPS2021, Ramped Load on Enabled Device

Figure 9. TPS2022, Ramped Load on Enabled Device

PARAMETER MEASUREMENT INFORMATION
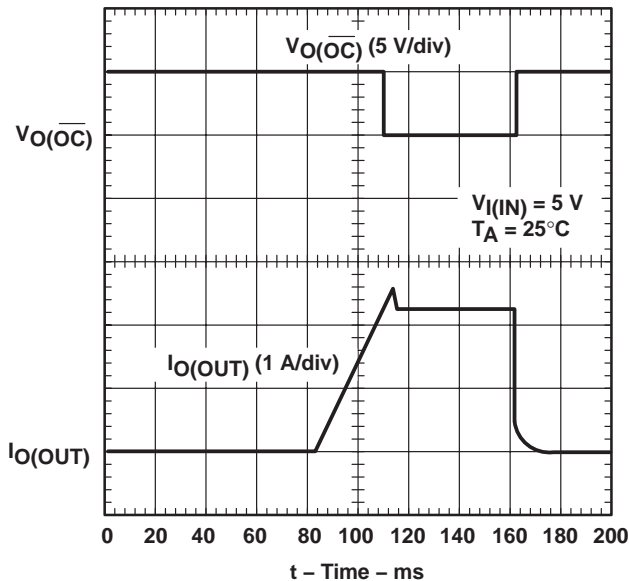
Figure 10. TPS2023, Ramped Load on
Enabled Device
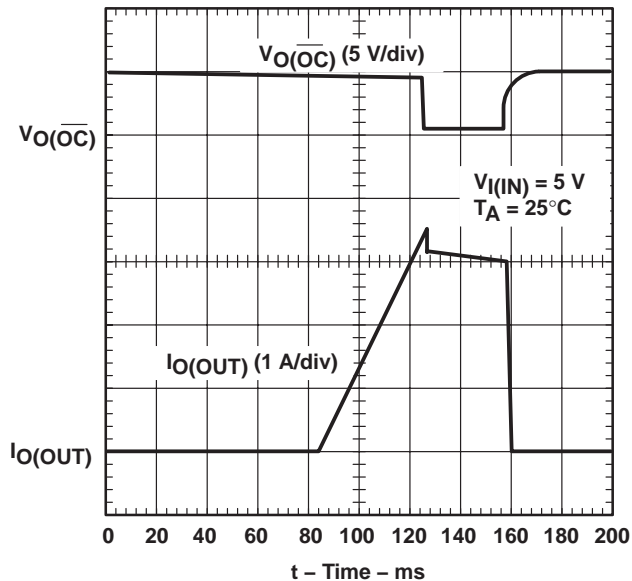
Figure 11. TPS2024, Ramped Load on Enabled
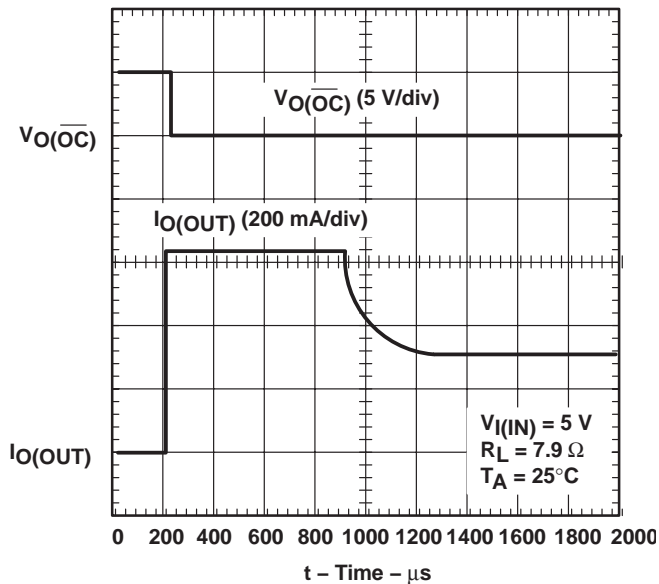Device
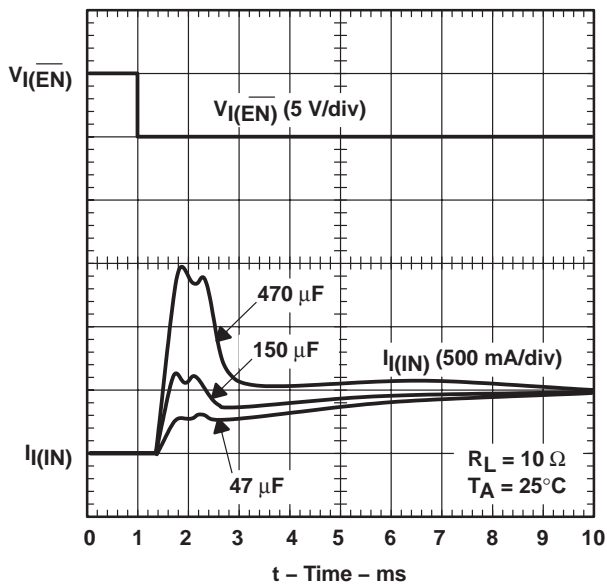
Figure 12. TPS2024, Inrush Current

Figure 13. 7.9-Ω Load Connected to an Enabled
TPS2020 Device

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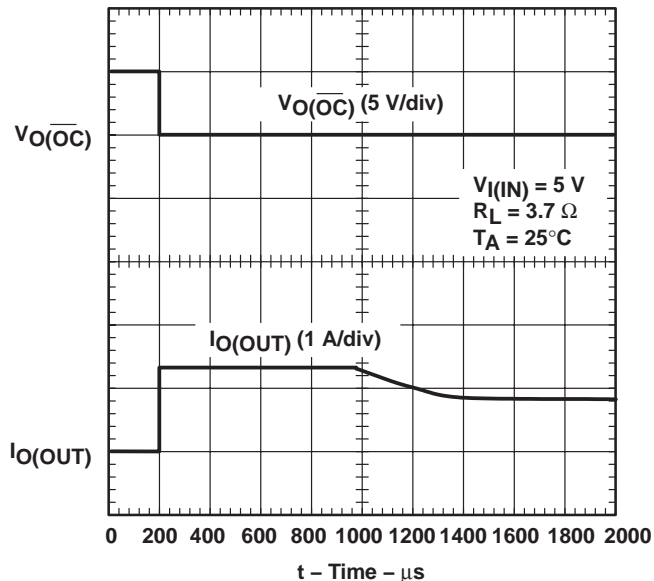INSTRUMENTS
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PARAMETER MEASUREMENT INFORMATION
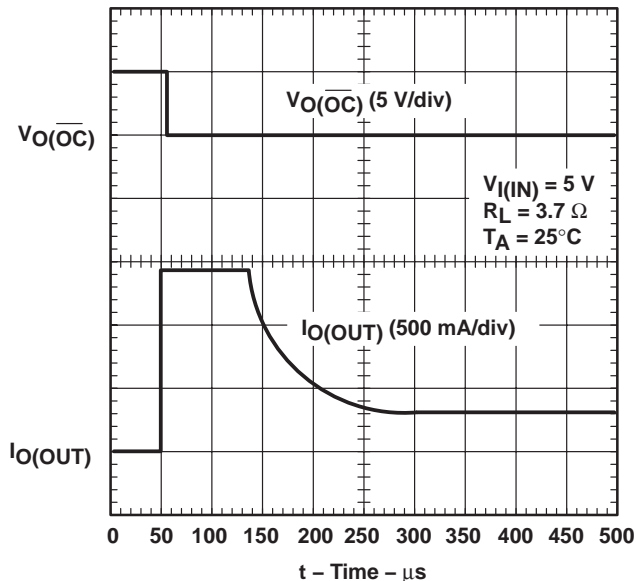
Figure 14. 3.7-Ω Load Connected to an Enabled TPS2020 Device

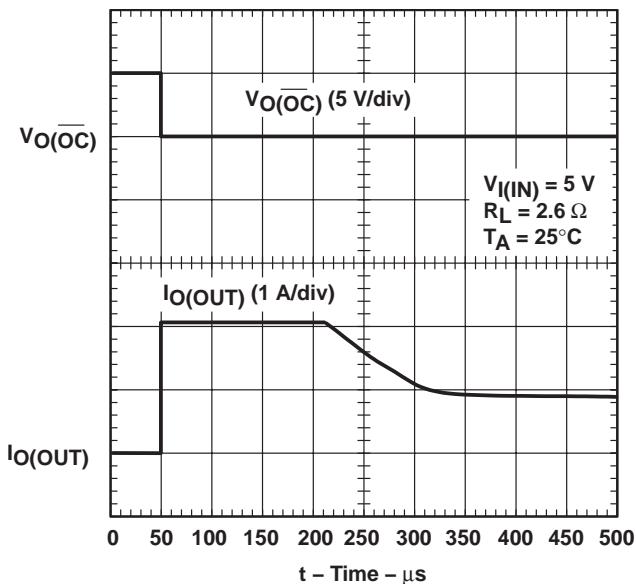**Figure 15. 3.7-Ω Load Connected to an Enabled TPS2021 Device**

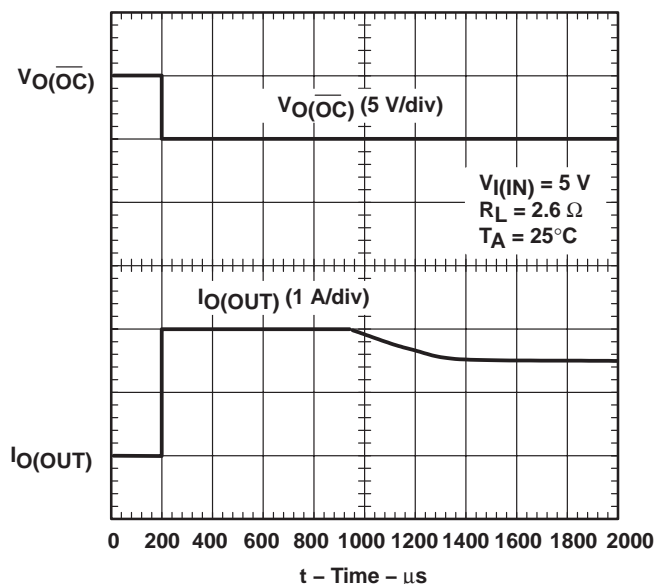**Figure 16. 2.6-Ω Load Connected to an Enabled TPS2021 Device**

Figure 17. 2.6-Ω Load Connected to an Enabled TPS2022 Device
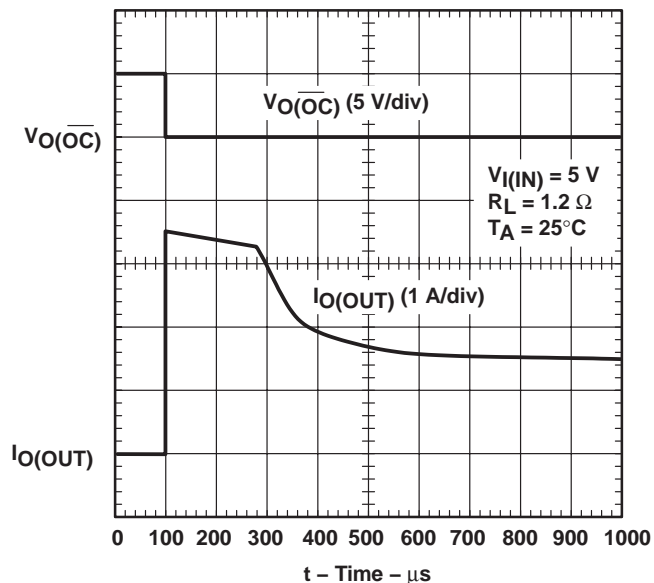
PARAMETER MEASUREMENT INFORMATION

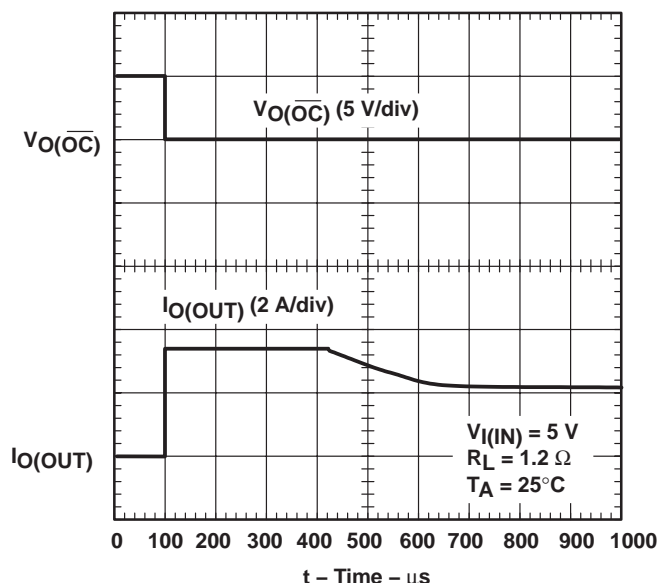**Figure 18. 1.2-Ω Load Connected to an Enabled TPS2022 Device**

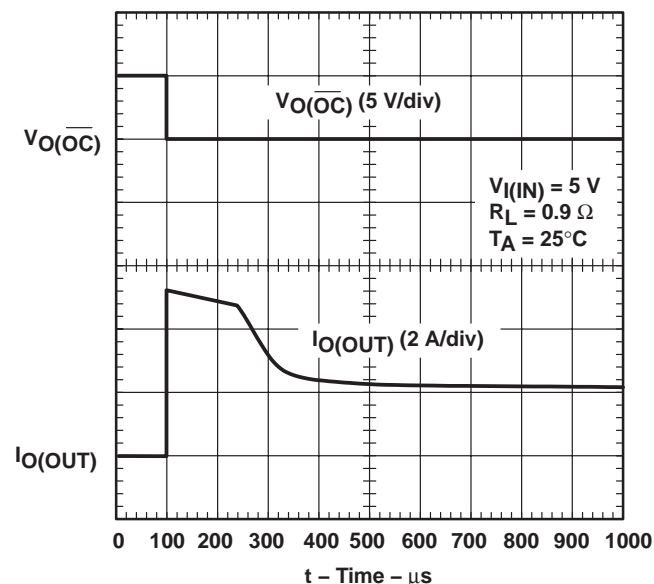**Figure 19. 1.2-Ω Load Connected to an Enabled TPS2023 Device**

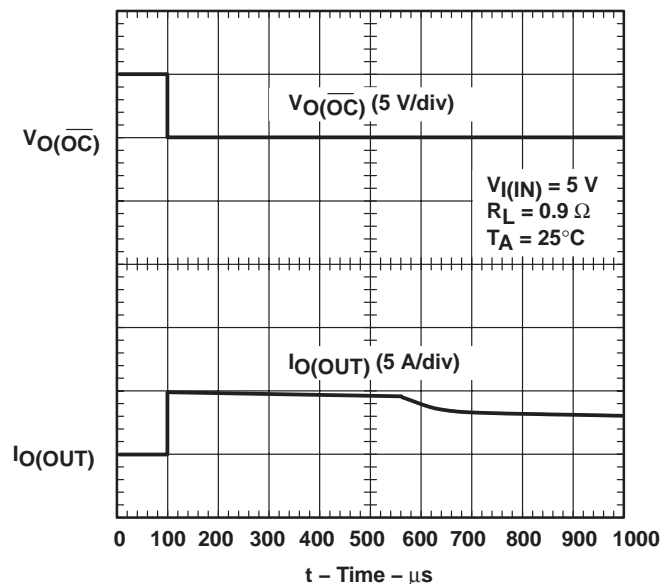**Figure 20. 0.9-Ω Load Connected to an Enabled TPS2023 Device**

Figure 21. 0.9-Ω Load Connected to an Enabled TPS2024 Device
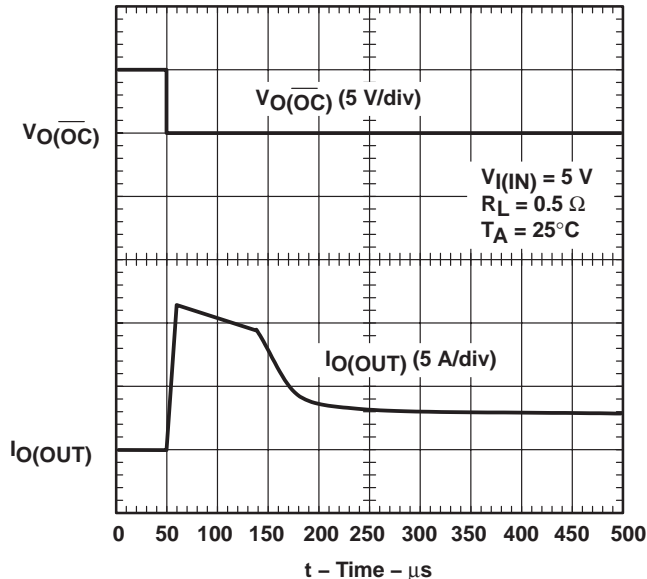
PARAMETER MEASUREMENT INFORMATION

$V_{I(IN)} = 5\ V$
$R_L = 0.5\ \Omega$
$T_A = 25°C$

Figure 22. 0.5-Ω Load Connected to an Enabled
TPS2024 Device

TYPICAL CHARACTERISTICS

Table of Graphs

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t_f	Fall time	vs Load current	26
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	Supply current (disabled)	vs Input voltage	30
I_{OS}	Short-circuit current limit	vs Input voltage	31
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$r_{DS(on)}$	Static drain-source on-state resistance	vs Input voltage	33
		vs Junction temperature	34
		vs Input voltage	35
		vs Junction temperature	36
V_I	Input voltage	Undervoltage lockout	37

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TYPICAL CHARACTERISTICS

TURNON DELAY TIME
vs
OUTPUT VOLTAGE

$T_A = 25°C$
$C_L = 1 \mu F$
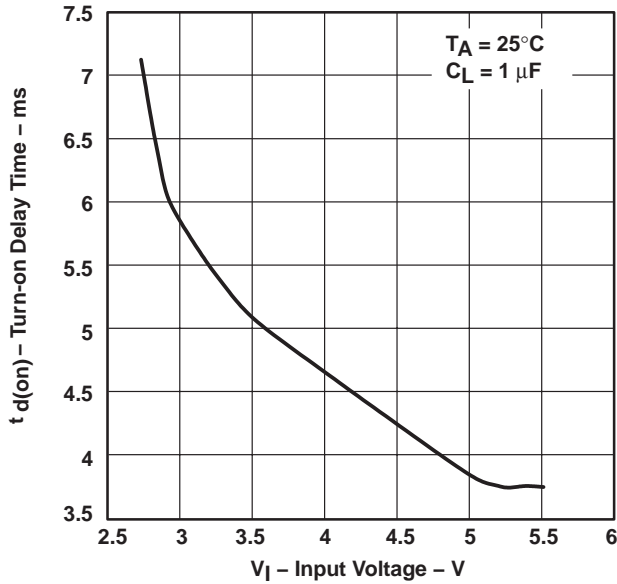
Figure 23

TURNOFF DELAY TIME
vs
INPUT VOLTAGE

$T_A = 25°C$
$C_L = 1 \mu F$
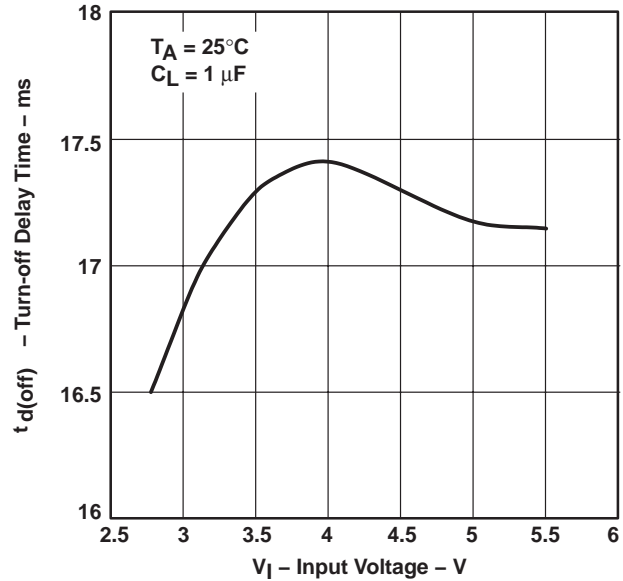
Figure 24

RISE TIME
vs
LOAD CURRENT

$T_A = 25°C$
$C_L = 1 \mu F$
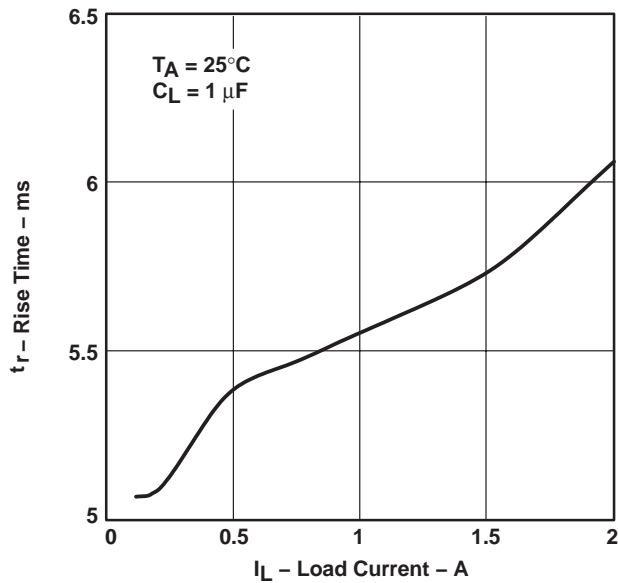
Figure 25

FALL TIME
vs
LOAD CURRENT

$T_A = 25°C$
$C_L = 1 \mu F$

Figure 26

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TYPICAL CHARACTERISTICS

SUPPLY CURRENT (ENABLED)
vs
JUNCTION TEMPERATURE

Figure 27

SUPPLY CURRENT (DISABLED)
vs
JUNCTION TEMPERATURE

Figure 28

SUPPLY CURRENT (ENABLED)
vs
INPUT VOLTAGE

Figure 29

SUPPLY CURRENT (DISABLED)
vs
INPUT VOLTAGE
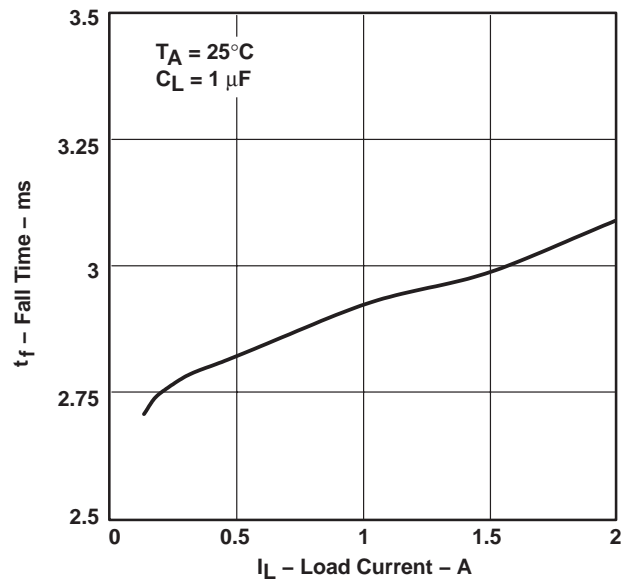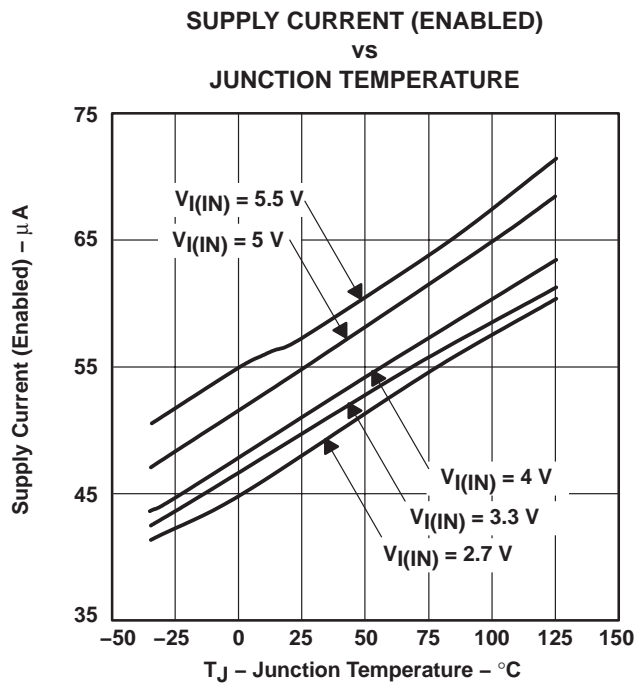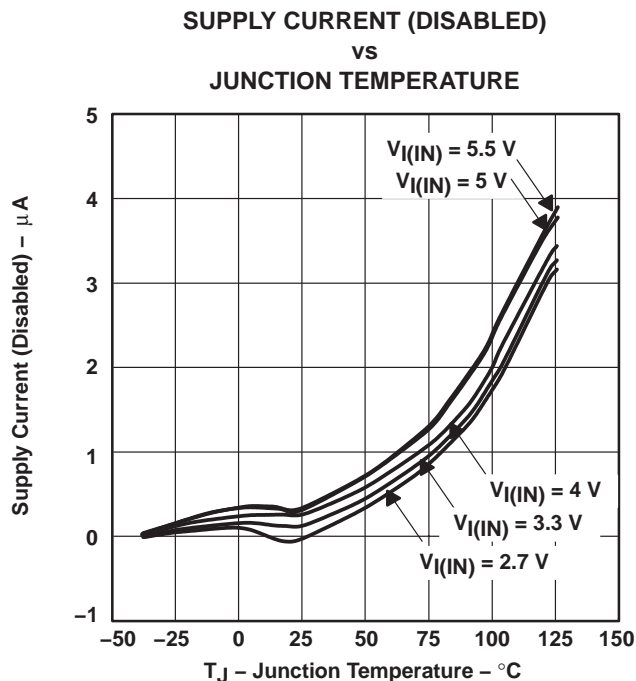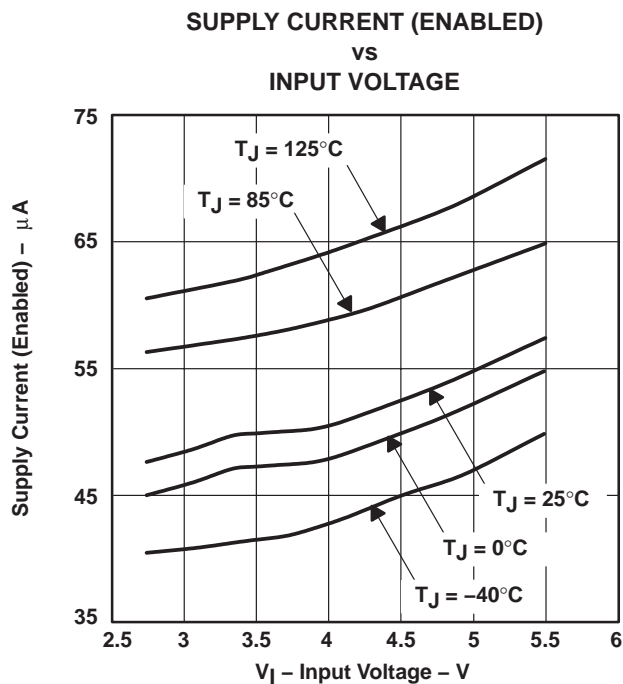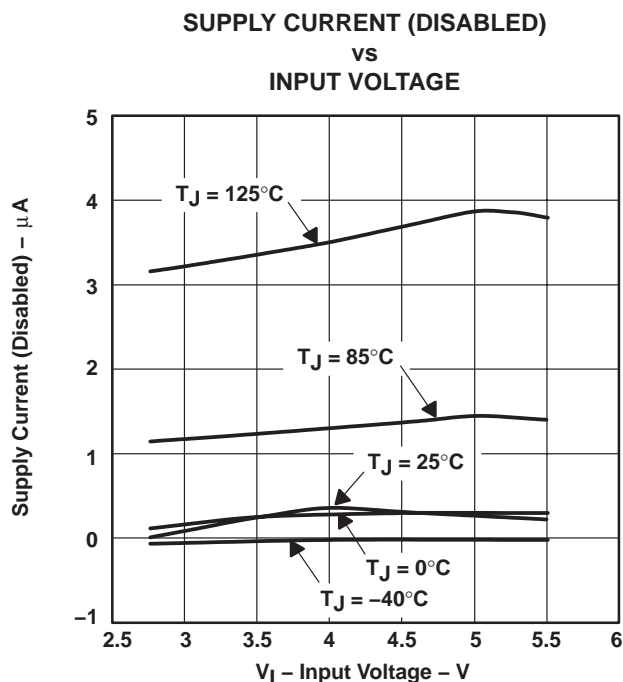
Figure 30

TYPICAL CHARACTERISTICS

SHORT-CIRCUIT CURRENT LIMIT
vs
INPUT VOLTAGE
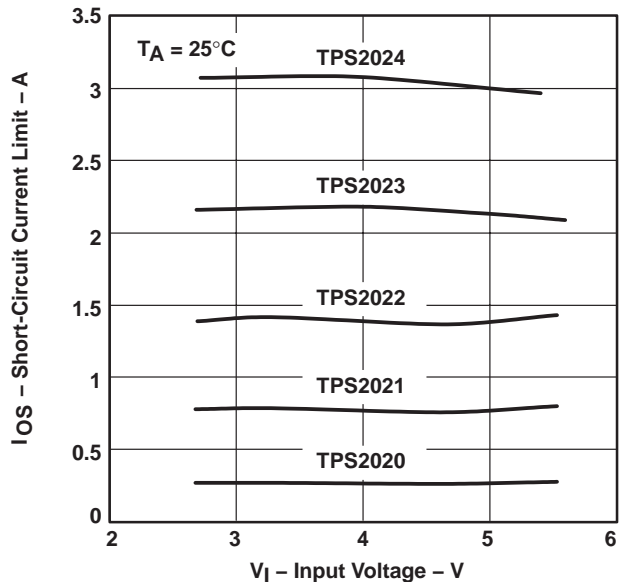
Figure 31

SHORT-CIRCUIT CURRENT LIMIT
vs
JUNCTION TEMPERATURE
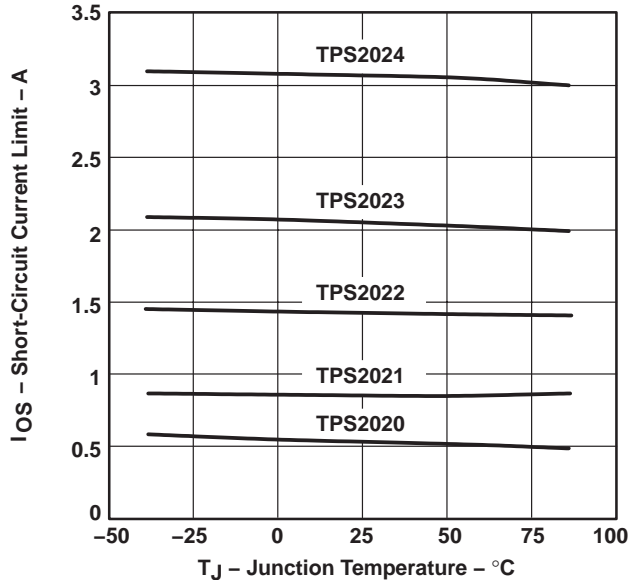
Figure 32

STATIC DRAIN-SOURCE ON-STATE RESISTANCE
vs
INPUT VOLTAGE
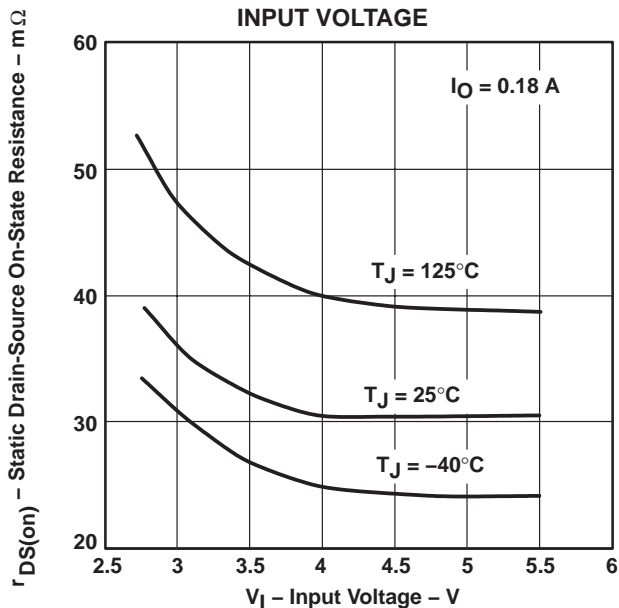
Figure 33

STATIC DRAIN-SOURCE ON-STATE RESISTANCE
vs
JUNCTION TEMPERATURE

Figure 34

TEXAS
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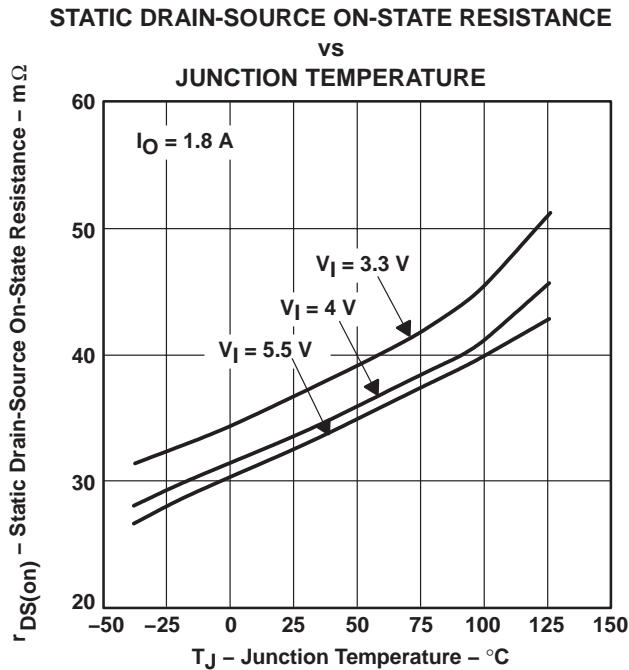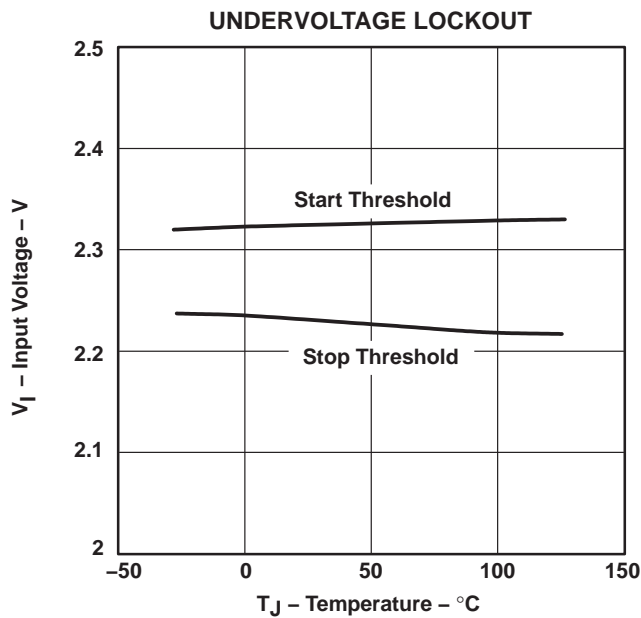
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TYPICAL CHARACTERISTICS

STATIC DRAIN-SOURCE ON-STATE RESISTANCE
vs
INPUT VOLTAGE

Figure 35

STATIC DRAIN-SOURCE ON-STATE RESISTANCE
vs
JUNCTION TEMPERATURE

Figure 36

UNDERVOLTAGE LOCKOUT

Figure 37

APPLICATION INFORMATION
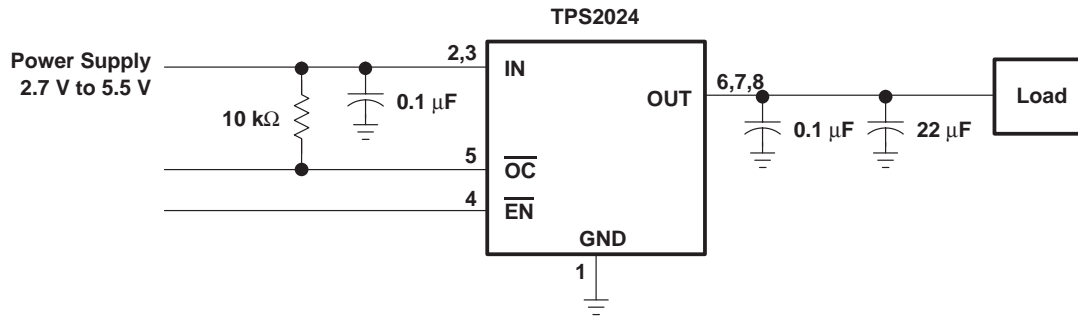
Figure 38. Typical Application

power-supply considerations

A 0.01-µF to 0.1-µF ceramic bypass capacitor between IN and GND, close to the device, is recommended. Placing a high-value electrolytic capacitor on the output and input pins is recommended when the output load is heavy. This precaution reduces power supply transients that may cause ringing on the input. Additionally, bypassing the output with a 0.01-µF to 0.1-µF ceramic capacitor improves the immunity of the device to short-circuit transients.

overcurrent

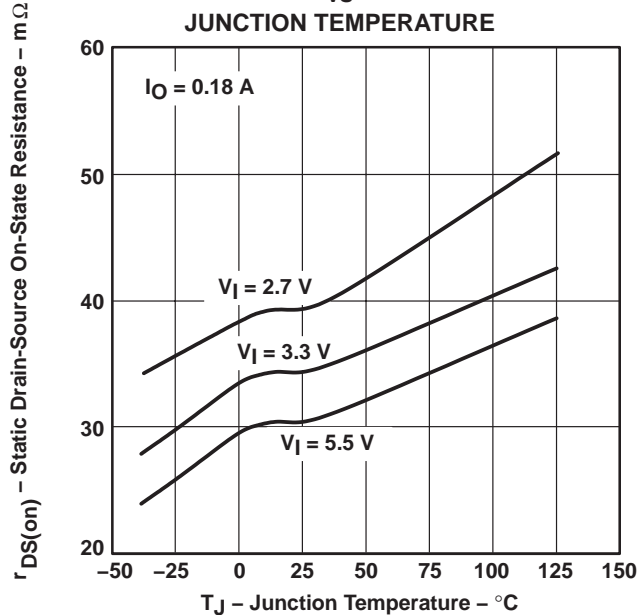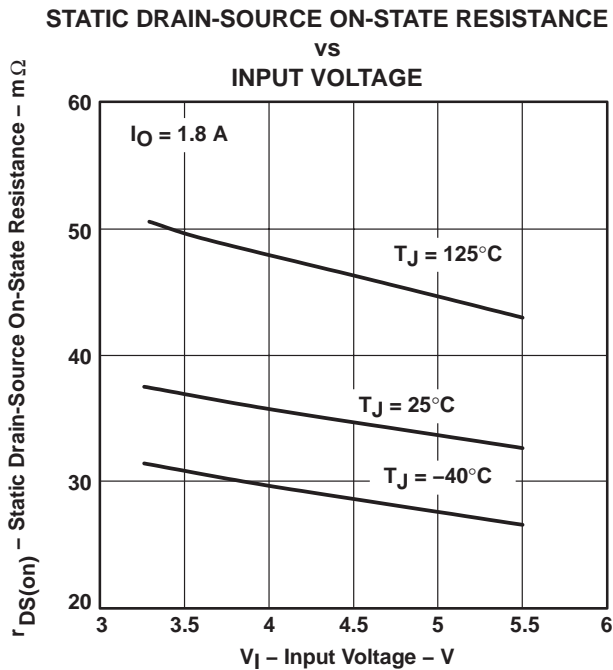A sense FET checks for overcurrent conditions. Unlike current-sense resistors, sense FETs do not increase the series resistance of the current path. When an overcurrent condition is detected, the device maintains a constant output current and reduces the output voltage accordingly. Complete shutdown occurs only if the fault is present long enough to activate thermal limiting.

Three possible overload conditions can occur. In the first condition, the output has been shorted before the device is enabled or before $V_{I(IN)}$ has been applied (see Figure 6). The TPS202x senses the short and immediately switches into a constant-current output.

In the second condition, the excessive load occurs while the device is enabled. At the instant the excessive load occurs, very high currents may flow for a short time before the current-limit circuit can react (see Figures 13–22). After the current-limit circuit has tripped (reached the overcurrent trip threshhold) the device switches into constant-current mode.

In the third condition, the load has been gradually increased beyond the recommended operating current. The current is permitted to rise until the current-limit threshold is reached or until the thermal limit of the device is exceeded (see Figures 7–11). The TPS202x is capable of delivering current up to the current-limit threshold without damaging the device. Once the threshold has been reached, the device switches into its constant-current mode.

$\overline{OC}$ response

The $\overline{OC}$ open-drain output is asserted (active low) when an overcurrent or overtemperature condition is encountered. The output will remain asserted until the overcurrent or overtemperature condition is removed. Connecting a heavy capacitive load to an enabled device can cause momentary false overcurrent reporting from the inrush current flowing through the device, charging the downstream capacitor. An RC filter can be connected to the $\overline{OC}$ pin to reduce false overcurrent reporting. Using low-ESR electrolytic capacitors on the output lowers the inrush current flow through the device during hot-plug events by providing a low impedance energy source, thereby reducing erroneous overcurrent reporting.
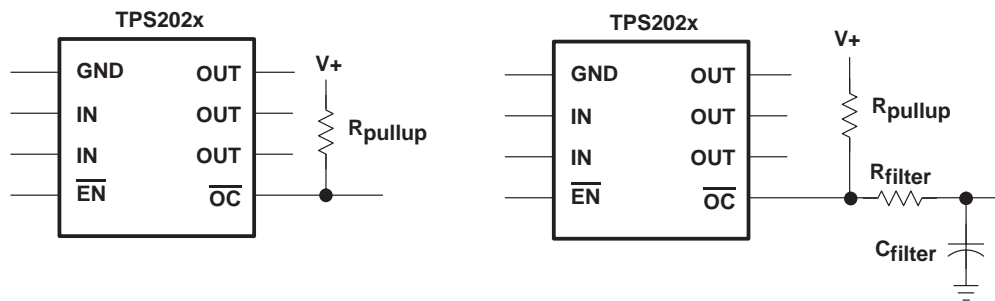
APPLICATION INFORMATION

Figure 39. Typical Circuit for $\overline{OC}$ Pin and RC Filter for Damping Inrush $\overline{OC}$ Responses

power dissipation and junction temperature

The low on-resistance on the n-channel MOSFET allows small surface-mount packages, such as SOIC, to pass large currents. The thermal resistances of these packages are high compared to those of power packages; it is good design practice to check power dissipation and junction temperature. The first step is to find $r_{DS(on)}$ at the input voltage and operating temperature. As an initial estimate, use the highest operating ambient temperature of interest and read $r_{DS(on)}$ from Figures 33–36. Next, calculate the power dissipation using:

$$P_D = r_{DS(on)} \times I^2$$

Finally, calculate the junction temperature:

$$T_J = P_D \times R_{\theta JA} + T_A$$

Where:

T_A = Ambient Temperature °C
$R_{\theta JA}$ = Thermal resistance SOIC = 172°C/W, PDIP = 106°C/W

Compare the calculated junction temperature with the initial estimate. If they do not agree within a few degrees, repeat the calculation, using the calculated value as the new estimate. Two or three iterations are generally sufficient to get an acceptable answer.

thermal protection

Thermal protection prevents damage to the IC when heavy-overload or short-circuit faults are present for extended periods of time. The faults force the TPS202x into constant current mode, which causes the voltage across the high-side switch to increase; under short-circuit conditions, the voltage across the switch is equal to the input voltage. The increased dissipation causes the junction temperature to rise to high levels. The protection circuit senses the junction temperature of the switch and shuts it off. Hysteresis is built into the thermal sense circuit, and after the device has cooled approximately 20 degrees, the switch turns back on. The switch continues to cycle in this manner until the load fault or input power is removed.

undervoltage lockout (UVLO)

An undervoltage lockout ensures that the power switch is in the off state at powerup. Whenever the input voltage falls below approximately 2 V, the power switch will be quickly turned off. This facilitates the design of hot-insertion systems where it is not possible to turn off the power switch before input power is removed. The UVLO will also keep the switch from being turned on until the power supply has reached at least 2 V, even if the switch is enabled. Upon reinsertion, the power switch will be turned on, with a controlled rise time to reduce EMI and voltage overshoots.

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APPLICATION INFORMATION

generic hot-plug applications (see Figure 40)

In many applications it may be necessary to remove modules or pc boards while the main unit is still operating. These are considered hot-plug applications. Such implementations require the control of current surges seen by the main power supply and the card being inserted. The most effective way to control these surges is to limit and slowly ramp the current and voltage being applied to the card, similar to the way in which a power supply normally turns on. Because of the controlled rise times and fall times of the TPS202x series, these devices can be used to provide a softer start-up to devices being hot-plugged into a powered system. The UVLO feature of the TPS202x also ensures the switch will be off after the card has been removed, and the switch will be off during the next insertion. The UVLO feature guarantees a soft start with a controlled rise time for every insertion of the card or module.
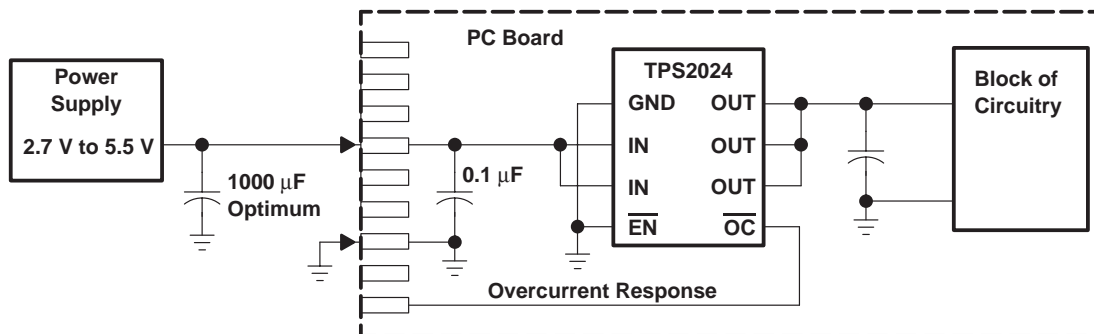
Figure 40. Typical Hot-Plug Implementation

By placing the TPS202x between the V_{CC} input and the rest of the circuitry, the input power will reach this device first after insertion. The typical rise time of the switch is approximately 9 ms, providing a slow voltage ramp at the output of the device. This implementation controls system surge currents and provides a hot-plugging mechanism for any device.

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PACKAGING INFORMATION

Orderable Device	Status [1]	Package Type	Package Drawing	Pins	Package Qty	Eco Plan [2]	Lead/Ball Finish	MSL Peak Temp [3]
TPS2020D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2020DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2020DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2020P	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2020PE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2021D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2021DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2021DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2021DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2021P	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2021PE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2022D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2022DG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2022DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2022DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2022P	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2022PE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2023D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2023DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2023P	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2023PE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC
TPS2024D	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2024DR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2024DRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM
TPS2024P	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC

Orderable Device	Status [1]	Package Type	Package Drawing	Pins	Package Qty	Eco Plan [2]	Lead/Ball Finish	MSL Peak Temp [3]
TPS2024PE4	ACTIVE	PDIP	P	8	50	Pb-Free (RoHS)	CU NIPDAU	Level-NC-NC-NC

[1] The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

[2] Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.
TBD: The Pb-Free/Green conversion plan has not been defined.
Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.
Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

[3] MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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P (R-PDIP-T8)   **PLASTIC DUAL-IN-LINE**

0.400 (10,60)
0.355 (9,02)

8   5

0.260 (6,60)
0.240 (6,10)

1   4

0.070 (1,78) MAX

0.020 (0,51) MIN

0.325 (8,26)
0.300 (7,62)

0.015 (0,38)

Gage Plane

0.200 (5,08) MAX

Seating Plane

0.010 (0,25) NOM

0.125 (3,18) MIN

0.100 (2,54)

0.430 (10,92)
MAX

0.021 (0,53)
0.015 (0,38) $\oplus$ 0.010 (0,25) Ⓜ
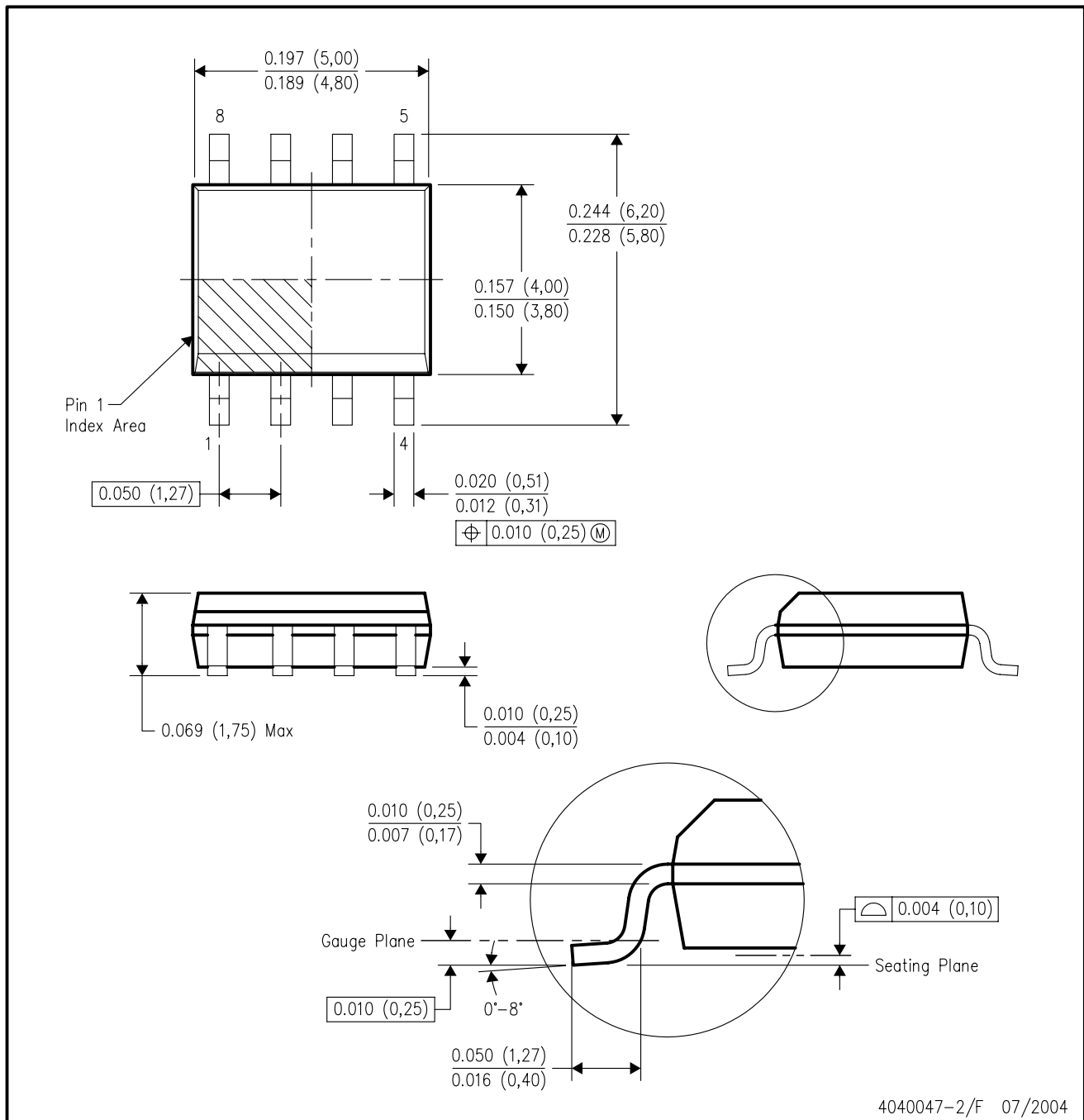
4040082/D 05/98

NOTES: A.   All linear dimensions are in inches (millimeters).
         B.   This drawing is subject to change without notice.
         C.   Falls within JEDEC MS-001

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D (R−PDSO−G8)   PLASTIC SMALL−OUTLINE PACKAGE

0.197 (5,00)
0.189 (4,80)

8   5

0.244 (6,20)
0.228 (5,80)

0.157 (4,00)
0.150 (3,80)

Pin 1
Index Area

1   4

0.050 (1,27)

0.020 (0,51)
0.012 (0,31)

⊕ 0.010 (0,25) Ⓜ

0.069 (1,75) Max

0.010 (0,25)
0.004 (0,10)

0.010 (0,25)
0.007 (0,17)

0.004 (0,10)

Gauge Plane

Seating Plane

0.010 (0,25)

0°−8°

0.050 (1,27)
0.016 (0,40)

4040047−2/F   07/2004

NOTES: A.   All linear dimensions are in inches (millimeters).
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         C.   Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
         D.   Falls within JEDEC MS−012 variation AA.

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