## Constant－current driver

## for small LCD panel back light

## （Non－boost type） BD1754HFN

## －General Description

The multi－level brightness control LED works as a constant current driver in 64 steps，so that the driving current can be adjusted finely．BD1754HFN is best suited to turn on LEDs that require high－accuracy LED brightness control．

## －Features

1）Current regulation for LED up to 4 parallels
2）Adjustable constant current 64 steps
3）High accurate and matching of each current channel（0．5\％Typ）
4）Brightness control via a signal－line digital control interface（Uni－Port Interface Control＝UPIC）
5） $2.9 \mathrm{~mm} \times 3.0 \mathrm{~mm}$ HSON8 Small package
－Application
This driver is applicable for various fields such as mobile phones，portable game machines and etc．

- Absolute Maximum Ratings ( $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Limits | Unit |
| :--- | :---: | :---: | :---: |
| Maximum applied voltage | VMAX | 7 | V |
| Power dissipation | Pd | $630\left({ }^{*} 1\right)$ | mW |
| Operating temperature range | Topr | $-30 \sim+85$ | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature range | Tstg | $-55 \sim+150$ | ${ }^{\circ} \mathrm{C}$ |

(*1) This value is the measurement value when the driver is mounted on a glass epoxy board ( $70 \mathrm{~mm} \times 70 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ ). When using the driver at Ta of $25^{\circ} \mathrm{C}$ or higher, the power is dissipated by approx. $5.04 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$.

- Recommended Operating Conditions ( $\mathrm{Ta}=-30^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ )

| Parameter |  | Symbol | Limits |  |  | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |

- Electrical Characteristics
(Unless otherwise specified, $\mathrm{Ta}=25^{\circ} \mathrm{C}$ and Vin $=3.6 \mathrm{~V}$ )

| Parameter | Symbol | Limits |  |  | Parameter | Symbol |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Typ | Max |  |  |
| Quiescent current | 19 | - | 0.1 | 1 | $\mu \mathrm{A}$ | EN=0V |
| Circuit current | IDD | - | 1.2 | 2.0 | mA | Except LED current |
| [Current driver] |  |  |  |  |  |  |
| Maximum current | ILED-max | 29.76 | 32.0 | 34.24 | mA | $\mathrm{R}_{\text {ISET }}=120 \mathrm{k} \Omega$ |
| LED Current accuracy | ILED-diff | - | - | 7.0 | \% | When current 16.5 mA setting $\mathrm{R}_{\text {ISET }}=120 \mathrm{k} \Omega$ |
| LED Current matching | ILED-match | - | 0.5 | 3.0*1 | \% | When current 16.5 mA setting $R_{\text {ISET }}=120 \mathrm{k} \Omega$ |
| [Logic controller] |  |  |  |  |  |  |
| Low threshold voltage | VIL | - | - | 0.4 | V |  |
| High threshold voltage | VIH | 1.4 | - | - | V |  |
| 'H' level input current | IIH | - | 0 | 2 | $\mu \mathrm{A}$ | $\mathrm{EN}=\mathrm{Vin}$ |
| 'L' level input current | IIL | -2 | 0 | - | $\mu \mathrm{A}$ | EN=0V |
| EN 'H' time | THI | 0.05 | - | 100 | $\mu \mathrm{sec}$ |  |
| EN 'L' time | TLO | 0.3 | - | 100 | $\mu \mathrm{sec}$ |  |
| EN Off time-out | TOFF | 1 | - | - | msec |  |
| VIN supply -> EN active time | TVINON | 1 | - | - | msec |  |
| EN stand-by -> VBAT Off time | TVINOFF | 0 | - | - | msec |  |

*1) The following expression is used for calculation:
$I_{\text {LED-match }}=\{(\operatorname{Imax}-\operatorname{Imin}) /(I \max +\operatorname{Imin})\} \times 100$
Imax = Current value in a channel with the maximum current value among all channels
Imin = Current value in a channel with the minimum current value among all channels

## - Reference Data



Fig. 1 Circuit current (stand-by)


Fig. 2 Circuit current


Fig. 3 LED off-leakage current


L1 Terminal Voltage [V]
Fig. 4 LED output current vs. LED pin voltage (Vin $=3.6 \mathrm{~V}$, at 32 mA of LED current)


Fig. 7 LED current characteristics
(Vin $=3.6 \mathrm{~V}$, integral linearity error)


Fig. 5 LED output current vs. $\mathrm{V}_{\mathbb{N}}$
(Vin $=3.6 \mathrm{~V}$, at 32 mA of LED current)


Fig. 8 LED current relative accuracy


Fig. 6 LED current characteristics (Vin $=3.6 \mathrm{~V}$, differential linearity error)


Fig. 9 LED current vs. R RIST

$$
(\mathrm{Vin}=3.6 \mathrm{~V})
$$

(Vin $=3.6 \mathrm{~V}$, at the maximum current setting)


Fig. 10 BD1754HFN Block Diagram and Recommended Circuit Example

- Terminals

| No. | Pin Name | $\begin{aligned} & \text { In/ } \\ & \text { Out } \end{aligned}$ | ESD Diode |  | Functions |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | For Power | For GND |  |
| 1 | EN | In | VIN | GND | LED enable and Brightness control signal |
| 2 | GND | - | VIN | - | Ground |
| 3 | ISET | Out | VIN | GND | Bias current |
| 4 | VIN | - | - | GND | Power supply voltage input |
| 5 | L1 | Out | - | GND | Current sink for LED 1 |
| 6 | L2 | Out | - | GND | Current sink for LED 2 |
| 7 | L3 | Out | - | GND | Current sink for LED 3 |
| 8 | L4 | Out | - | GND | Current sink for LED 4 |

- Description of Operations
(1) UPIC (= Uni-Port Interface Control) interface

BD1754HFN has a single-line digital control interface (UPIC) that can control the power ON/OFF and LED current value through the EN pin only. The LED current decreases by one step depending on the number of rising edges. After the number of rising edge is reached to the minimum output current ( 64 rising edges), the next rising edge changes the output current to the maximum value at startup time. To maintain any output current, the EN pin must be kept at ' H ' level. To power off, the EN pin must be kept at ' L ' level for more than 1 msec .


Fig. 11 Brightness Control Method


Fig. 12 UPIC Interface

By following sequence, UPIC can control current driver for MAX current and OFF state only.


Fig. 13 UPIC Interface usage for MAX current or OFF only
(2) Current Driver

The MAX Current is determined by the ISET resistance and the following expression.

$$
\mathrm{ILED}-\mathrm{MAX}[\mathrm{~mA}]=6.4 \times 600[\mathrm{mV}] / \mathrm{R}_{\text {ISET }}[\mathrm{k} \Omega]
$$

The LED current state can be changed by the EN control signal. When the state is Cn, the output current (led) can be obtained from the following expression (where, n indicates a state number).

$$
I_{\text {LED }}[\mathrm{mA}]=\mathrm{I}_{\text {LED-MAX }} \times \mathrm{n} / 64
$$

The following table is the example of LED current, when ISET resistance is 120 [ $\mathrm{k} \Omega$ ].
$\mathrm{R}_{\text {ISET }}: 120[\mathrm{k} \Omega]$

| State | Output current <br> $[\mathrm{mA}]$ | State | Output current <br> $[\mathrm{mA}]$ | State | Output current <br> $[\mathrm{mA}]$ | State | Output current <br> $[\mathrm{mA}]$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| C64 | 32.0 | C48 | 24.0 | C32 | 16.0 | C16 | 8.0 |
| C63 | 31.5 | C47 | 23.5 | C31 | 15.5 | C15 | 7.5 |
| C62 | 31.0 | C46 | 23.0 | C30 | 15.0 | C14 | 7.0 |
| C61 | 30.5 | C45 | 22.5 | C29 | 14.5 | C13 | 6.5 |
| C60 | 30.0 | C44 | 22.0 | C28 | 14.0 | C12 | 6.0 |
| C59 | 29.5 | C43 | 21.5 | C27 | 13.5 | C11 | 5.5 |
| C58 | 29.0 | C42 | 21.0 | C26 | 13.0 | C10 | 5.0 |
| C57 | 28.5 | C41 | 20.5 | C25 | 12.5 | C9 | 4.5 |
| C56 | 28.0 | C40 | 20.0 | C24 | 12.0 | C8 | 4.0 |
| C55 | 27.5 | C39 | 19.5 | C23 | 11.5 | C7 | 3.5 |
| C54 | 27.0 | C38 | 19.0 | C22 | 11.0 | C6 | 3.0 |
| C53 | 26.5 | C37 | 18.5 | C21 | 10.5 | C5 | 2.5 |
| C52 | 26.0 | C36 | 18.0 | C20 | 10.0 | C4 | 2.0 |
| C51 | 25.5 | C35 | 17.5 | C19 | 9.5 | C3 | 1.5 |
| C50 | 25.0 | C34 | 17.0 | C18 | 9.0 | C2 | 1.0 |
| C49 | 24.5 | C33 | 16.5 | C17 | 8.5 | C1 | 0.5 |

When the state is C 64 (the maximum value), the output current value can be changed on the ISET resistance value as below.

State : C64

| ISET resistance value $(k \Omega)$ | Output current per channel <br> $(\mathrm{mA})$ | Total output current of the four <br> channels $(\mathrm{mA})$ |
| :---: | :---: | :---: |
| 240 | 16.0 | 64.0 |
| 120 | 32.0 | 128.0 |
| 90 | 42.7 | 170.8 |
| 60 | 64.0 | 256.0 |

## - Application Circuit Examples

(1) Circuit example when the power supply is separated


Fig. 14 Circuit example when the power supply is separated
This figure shows a circuit example when the power supply for VIN and for LEDs is separated. Apply a voltage of Vf (threshold voltage value of a white LED) or higher to the LED.
In this time, it is necessary to note that, when the LED is powered ON, the voltage value of $L^{*}$ pin (each pin of L1 to L4) must be $\mathrm{VIN}-1.4 \mathrm{~V}$ at the maximum. If a voltage of higher than $\mathrm{VIN}-1.4 \mathrm{~V}$ is applied to $\mathrm{L}^{*}$ pin, a desired current value cannot be obtained. Also, it is necessary to pay attention to the voltage application procedure at start-up. Be sure to power the current driver ON using the UPIC after applying power supply voltages to the VIN and the LED-anode pins. If the current driver is powered ON prior to applying power supply voltages to the LED, a rush current occurs in the LED.
Determine the resistance value with which the LED current value is maximized and then connect such resistor between the ISET and the GND pins. The power ON/OFF and the brightness of the LEDs are controlled through the EN pin in accordance with the UPIC format.
(2) Circuit example when using only two LEDs


Fig. 15 Circuit example when using only two LEDs

This figure shows a circuit example when both of the L3 and L4 LEDs are not used. Connect both of the unused L3 and L4 pins to the GND pin. Likewise, it is possible to make the L1 and/or the L2 pins unused, which allows the back lights to be used with the one or three LED(s) tumed on. In all cases, connect the unused $L^{*}$ pin to the GND pin.
Determine the resistance value with which the LED current value is maximized and then connect such resistor between the ISET and the GND pins. The power ON/OFF and the brightness of the LEDs are controlled through the EN pin in accordance with the UPIC format.


Fig. 16 Circuit example when the EN pin is powered on at all times

This figure shows a circuit example when the EN pin is powered on at all times. To prevent a rush current from occurring in the driver, it is necessary to apply voltages to the VIN pin and the LEDs prior to powering the current driver ON. Mount an RC filter between the VIN and the EN pins to delay the EN pin rising against the power-supply voltage rising. Determine the resistance value with which the LED current value is maximized and then connect such resistor between the ISET and the GND pins.
(4) Circuit example when performing a PWM brightness control


Fig. 17 Circuit example when performing a PWM brightness control

This figure shows a circuit example when performing a PWM brightness control. Through switching the ISET resistance value by the PWM input signal, the LED current is outputted under a PWM mode. The EN signal is controlled by an applied voltage level. In the circuit example shown above, the LED current value is changed to 3.43 mA in $0 \%$ of the PWM duty cycle, 17.72 mA in $50 \%$ of that and 32 mA in $100 \%$ of that.
(5) Circuit example when driving a large current with only one LED powered on.


Fig. 18 Circuit example when driving a large current with only one LED powered on.

This figure shows a circuit example when driving a large current through all the four channels with only one LED powered on. By shorting out all the LED driver pins, in the example of using $120 \mathrm{k} \Omega \mathrm{R}_{\text {ISET }}$, a current up to $128 \mathrm{~mA}(32 \mathrm{~mA} \times 4)$ can be driven. In this example, the brightness can be adjusted in 64 gradations with 2 mA step ( 0.5 mA step/channel $\times 4$ channels). For higher current values, using $60 \mathrm{k} \Omega$ R RISET allows a current up to 256 mA to be driven into one of the LEDs.
The power ON/OFF and the brightness of the LEDs are controlled through the EN pin in accordance with the UPIC format.
(6) Circuit example when making the eight LEDs available by connecting the two BD1754HFN drivers


Fig. 19 Circuit example when making the eight LEDs available by connecting the two BD1754HFN drivers

This figure shows a circuit example when making the eight LEDs available by connecting the two BD1754HFN drivers. By connecting the control signals to the EN pins in parallel, the eight LED channels can be controlled concurrently. This parallel connection can increase the number of the LED channels further as necessary (such as twelve, sixteen, or more).
Determine the resistance value with which the LED current value is maximized and then connect such resistor between the ISET and the GND pins.
The power ON/OFF and the brightness of the LEDs are controlled through the EN pin in accordance with the UPIC format.
(7) Circuit example when connecting the two LEDs to each of the channels in series


Fig. 20 Circuit example when connecting the two LEDs to each of the channels in series

This figure shows a circuit example when making the $8(2 \times 4)$ LEDs available by connecting the two LEDs to each of the channels in series. In this example, when $\mathrm{V} f$ is set to approx. 3 V in order to ensure the voltage to the L 1 to the L 4 pin, it is necessary to apply a voltage of 6.2 V ( 3 $\mathrm{V} \times 2$ LEDs in series +0.2 V of the minimum voltage value of the driver pin) or higher to the LED anode pin as its power supply voltage.
Pay attention that the voltage should not exceed the $7.0-\mathrm{V}$ maximum rating of the L 1 to the L 4 pin.
Determine the resistance value with which the LED current value is maximized and then connect such resistor between the ISET and the GND pins.
The power ON/OFF and the brightness of the LEDs are controlled through the EN pin in accordance with the UPIC format.

- Application Component Selection
<Capacitor>

| Symbol | Recommended value | Recommended component | Manufacturer |
| :---: | :---: | :---: | :---: |
| Cin | $0.1 \mu \mathrm{~F}$ | GRM188B31H104KA92B | MURATA |

<Resistor>

| Symbol | Recommended value | Recommended component | Manufacturer |
| :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\mathrm{ISET}}$ | $120 \mathrm{k} \Omega$ | MCR10PZHZF1203 | ROHM |

Recommended PCB Layout
Design PCB pattern to provide low impedance for the wiring to the power supply line.
Also, provide a bypass condenser if needed.

LED_PWR



Fig. 22 Surface (Top View)

Fig. 21 Layout image of the application components (Top View)

- Cautions on Use
(1) Absolute Maximum Ratings

An excess in the absolute maximum ratings, such as applied voltage, temperature range of operating conditions, can break down devices, thus making impossible to identify breaking mode such as a short circuit or an open circuit. If any special mode exceeding the absolute maximum ratings is assumed, consideration should be given to take physical safety measures including the use of fuses, etc.

## (2) Recommended Operating Conditions

These conditions represent a range within which characteristics can be provided approximately as expected. The electrical characteristics are guaranteed under the conditions of each parameter. The voltage and temperature characteristics are also shown under the conditions in respect of electrical ones.
(3) Reverse connection of power supply connector

The reverse connection of power supply connector can break down ICs. Take protective measures against the breakdown due to the reverse connection, such as mounting an external diode between the power supply and the IC's power supply terminal.
(4) Power supply line

Design PCB pattern to provide low impedance for the wiring between the power supply and the GND lines. Furthermore, for all power supply terminals to ICs, mount a capacitor between the power supply and the GND terminal. At the same time, in order to use an electrolytic capacitor, thoroughly check to be sure that the characteristics of the capacitor to be used present no problem including the occurrence of capacity dropout at a low temperature, thus determining the constant.
(5) GND voltage

Make setting of the potential of the GND terminal so that it will be maintained at the minimum in any operating state. Furthermore, check to be sure that no terminal is operated at a potential lower than the GND voltage including an actual electric transient.
(6) Short circuit between terminals and erroneous mounting

In order to mount ICs on a set PCB, pay thorough attention to the direction and offset of the ICs. Erroneous mounting can break down the ICs. Furthermore, if a short circuit occurs due to foreign matters entering between terminals or between the terminal and the power supply or the GND terminal, the ICs can break down.
(7) Operation in strong electromagnetic field

Be noted that using ICs in the strong electromagnetic field can cause a malfunction.
(8) Inspection with set PCB

On the inspection with the set PCB, if a capacitor is connected to a low-impedance IC terminal, the IC can suffer stress. Therefore, be sure to discharge from the set PCB by each process. Furthermore, in order to mount or dismount the set PCB to/from the jig for the inspection process, be sure to turn OFF the power supply and then mount the set PCB to the jig. After the completion of the inspection, be sure to turn OFF the power supply and then dismount it from the jig. In addition, for protection against static electricity, establish a ground for the assembly process and pay thorough attention to the transportation and the storage of the set PCB.
(9) Input terminals

In terms of the construction of IC, parasitic elements are inevitably formed in relation to potential. The operation of the parasitic element can cause interference with circuit operation, thus resulting in a malfunction and then breakdown of the input terminal. Therefore, pay thorough attention not to handle the input terminals, such as to apply to the input terminals a voltage lower than that applied to the GND respectively, so that any parasitic element will operate. Furthermore, do not apply a voltage to the input terminals when no power supply voltage is applied to the IC. In addition, even if the power supply voltage is applied, apply to the input terminals a voltage lower than the power supply voltage or within the guaranteed value of electrical characteristics.
(10) Ground wiring pattern

If small-signal GND and large-current GND are provided, it will be recommended to separate the large-current GND pattern from the small-signal GND pattern and establish a single ground at the reference point of the set PCB so that resistance to the wiring pattern and voltage fluctuations due to a large current will cause no fluctuations in voltages of the small-signal GND. Pay attention not to cause fluctuations in the GND wiring pattern of external parts as well.
(11) Thermal design

Perform thermal design in which there are adequate margins by taking into account the permissible dissipation (Pd) in actual states of use.

- Part Number Selection for Ordering


Part type


Part number


Package type HFN=HSON8


Taping type
TR = Embossed taping reel

## Taping

HSON8

<Package specification>

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